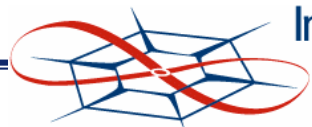


Network Analyzer Operation

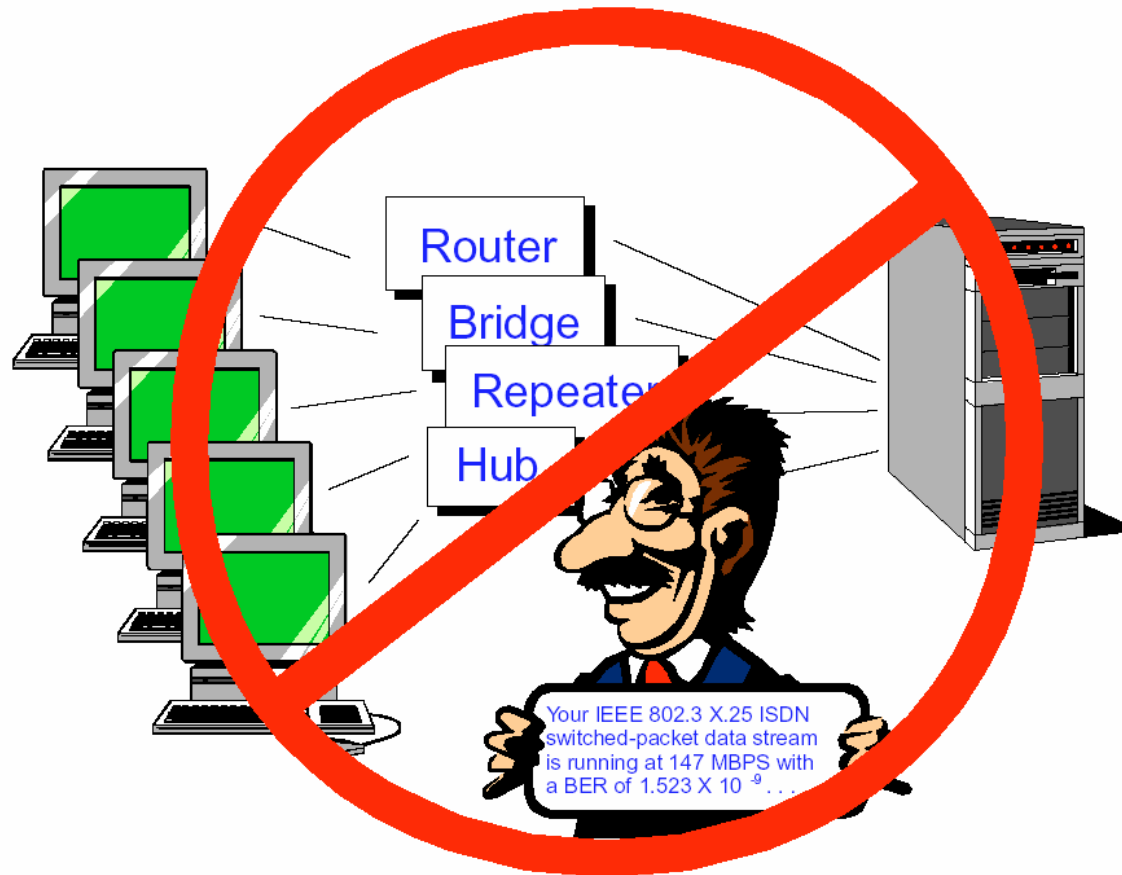
2004 ITTC Summer Lecture Series

John Paden



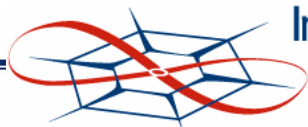
Purposes of a Network Analyzer

- Network analyzers are not about computer networks!



Purposes of a Network Analyzer

- Measures S-parameters of electronic devices.
 - E.g. Filters, amplifiers, mixers, switches, antennas, etc.
 - S-parameters are complex numbers (i.e. amplitude and phase)
- S-parameters are a generalization of the idea of the transfer function.
 - Remember transfer functions from EECS 360 (signal analysis course)
 - The transfer function of a device does not include information about the input and output impedance of a device. Therefore it does not tell you how the device will behave when connected to other components.
- Network analyzers are similar to continuous wave (CW) radar systems
 - These two systems share many features.



Terminology

- The device or system to be tested is referred to as the **DUT** or **Device Under Test**.
- The **test fixture** refers to the system outside of the network analyzer that is connected to the DUT.
 - While the test fixture is part of what the network analyzer measures, we ultimately want to measure the DUT by itself.
 - Most of the time, the test fixture is just a pair of cables used to connect the network analyzer to the device.
- **LTIV** stands for **Linear Time-Invariant**.

Overview of NA Operation

- The network analyzer measures in the frequency domain.
- The network analyzer **transmits** a sinusoid into the test fixture with a known frequency, amplitude, and phase.
- The network **receives** the amplitude and phase of one frequency from the output of the test fixture.
- The **transmitted** and **received** sinusoids do not have to be the same frequency.
 - **For most measurements the frequencies are the same.**

Linear Time-Invariant Devices

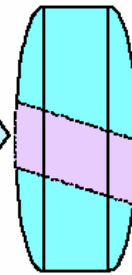
- A device is called LTIV if its operation can be explained by convolution.
- This means that the output signal is based on a infinite summation of scaled and time-delayed versions of the input signal.
 - The coefficients in this infinite summation never change.
- Now think about what happens when you have an infinite summation of scaled and time-delayed version of a single sinusoid.
- After all the summations you will end up with a sinusoid of the same frequency. Its phase and amplitude (which completely characterize it) are determined by the summations.

Network Analyzer Operation

$$R \cos(2\pi ft + \theta_R) = R e^{j\theta_R}$$

Incident
R

Reflected



* Incident, reflected, and transmitted fields are sinusoids

Transmitted
B

$$B \cos(2\pi ft + \theta_B) = B e^{j\theta_B}$$

$$A \cos(2\pi ft + \theta_A) = A e^{j\theta_A}$$

REFLECTION

$$\frac{\text{Reflected}}{\text{Incident}} = \frac{A}{R}$$

$$\frac{A}{R} e^{j\theta_A - \theta_R}$$

SWR

S-Parameters
S11, S22

Reflection
Coefficient
 Γ, ρ

Impedance,
Admittance
 $R+jX,$
 $G+jB$

Return
Loss

TRANSMISSION

$$\frac{\text{Transmitted}}{\text{Incident}} = \frac{B}{R}$$

$$\frac{B}{R} e^{j\theta_B - \theta_R}$$

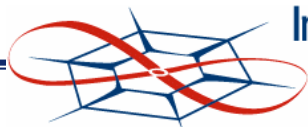
Gain / Loss

S-Parameters
S21, S12

Transmission
Coefficient
 T, τ

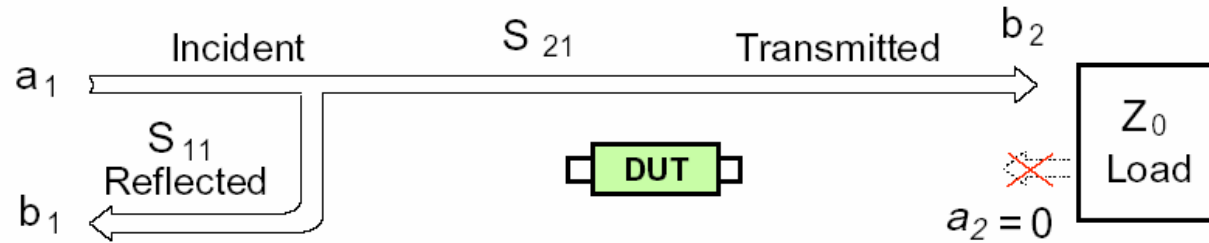
Insertion
Phase

Group
Delay



Network Analyzer Operation

Forward

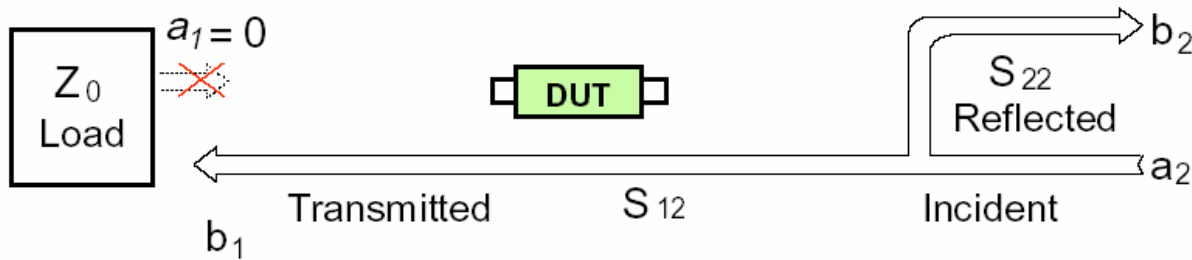


$$S_{11} = \frac{\text{Reflected}}{\text{Incident}} = \frac{b_1}{a_1} \Big|_{a_2 = 0}$$

$$S_{21} = \frac{\text{Transmitted}}{\text{Incident}} = \frac{b_2}{a_1} \Big|_{a_2 = 0}$$

$$S_{22} = \frac{\text{Reflected}}{\text{Incident}} = \frac{b_2}{a_2} \Big|_{a_1 = 0}$$

$$S_{12} = \frac{\text{Transmitted}}{\text{Incident}} = \frac{b_1}{a_2} \Big|_{a_1 = 0}$$



Reverse



Taking a Measurement: Start/Stop Freq

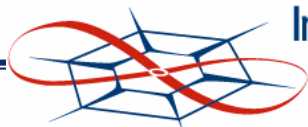
- Decide at which frequencies you want to know how your device performs.
 - Set your network analyzer's **start and stop frequencies**. If you plan to time-gate or process the data add some guard room on each side (i.e. make your start frequency a little lower and your stop frequency a little higher).
- PRISM SAR example: We have a Low Pass Filter (LPF) with a cutoff frequency of 90 MHz. We want to know what its passband behavior is and its stopband behavior at 450 to 470 MHz where the GPS telemetry radio link operates.
 - We will want to measure the device from near DC up to 500 MHz.
 - Suggest using HP 8753D (300 kHz-6 GHz)
 - start frequency: 300 kHz
 - stop frequency: 500 MHz



Taking a Measurement: # of points

- Determine the maximum length of the impulse response.
- PRISM SAR example: Continuing are LPF example, let us assume that we have a total of 2 meters of cable with a velocity factor of 69.5% and the LPF has four sections. We want to include up to 10 reflections through the system.

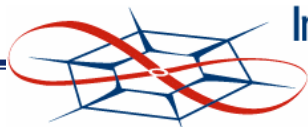
- Total time:
$$10 \left(\frac{2}{0.695 \cdot 3e8} + \frac{4}{90e6} \right) = 511 \text{ ns}$$



Taking a measurement: # of points

- We know the bandwidth (BW) of our measurement:
 - $BW = \text{stop frequency} - \text{start frequency} = 500 \text{ MHz}$
- We know the maximum length of the impulse response that we are interested in: 511 ns
- We can now calculate the **number of points** in the frequency domain we must sample at:

$$500 \text{ MHz} \cdot 511 \text{ ns} = \frac{511 \text{ ns}}{2 \text{ ns}} = 256 \text{ pts} \Rightarrow 401 \text{ pts}$$



Taking a measurement: Transmit Power

- We want to measure the stopband attenuation down to -90 dB with 16 dB SNR.
 - Our low pass filter is high-power so we can transmit at the highest power the network analyzer provides: 10 dBm **transmit power**.
 - The signal power with 90 dB of attenuation is then -80 dBm. To achieve the desired 16 dB SNR, the noise floor must be -96 dBm.
 - Using a noise figure of 53 dB for the network analyzer, the input noise power is:

$$P_{Noise} = 10\log_{10}(kTBF) + 30$$

K = Boltzmann's Constant ($1.38e-23$)

T = IEEE ref. Temp. (290 K)

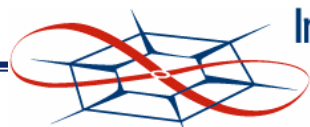
B = receiver bandwidth

F = Receiver noise figure (53 dB)

$$= 10\log_{10}(kT) + 10\log_{10}(B) + 10\log_{10}(F) + 30$$

$$= -204 + 10\log_{10}(B) + 53 + 30$$

$$= -121 + 10\log_{10}(B)$$



Taking a Measurement: IF Bandwidth

- To achieve an SNR of 16 dB, we need to set our receiver bandwidth so that $10\text{Log}_{10}(B)$ is 25 dB or less. Therefore B needs to be 300 Hz.
- The network analyzer calls receiver bandwidth “**IF bandwidth**” where IF stands for intermediate frequency.

Taking a Measurement: Averages

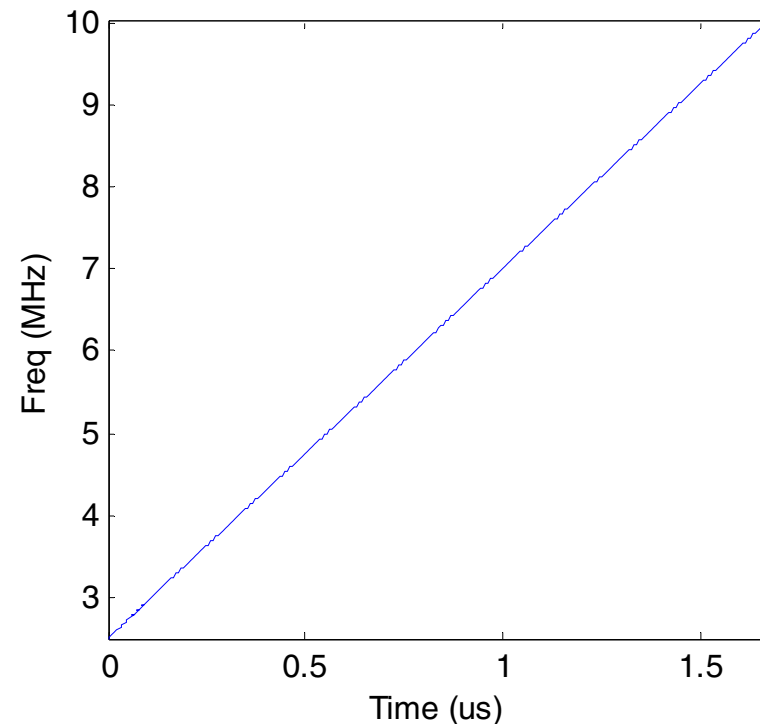
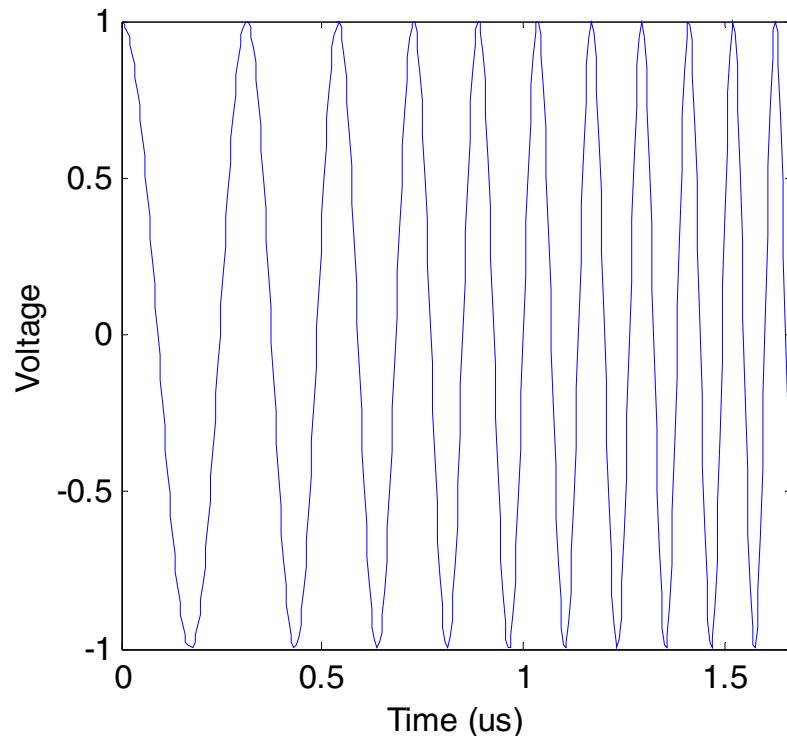
- Suppose we wanted to use an IF bandwidth of 10000 Hz (now $10\log_{10}(B) = 40$ dB).
- Another way to increase the SNR to the appropriate level would be to **average** 40 measurements. This effectively uses forty times the energy (similar to increasing the power by 16 dB).
- The whole equation becomes:

$$SNR = 10\log_{10}(P) + 10\log_{10}(N) - 10\log_{10}(L) - 10\log_{10}(kTBF)$$

where P = transmit power in Watts, N = number of averages, and L = loss of DUT.

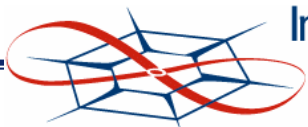
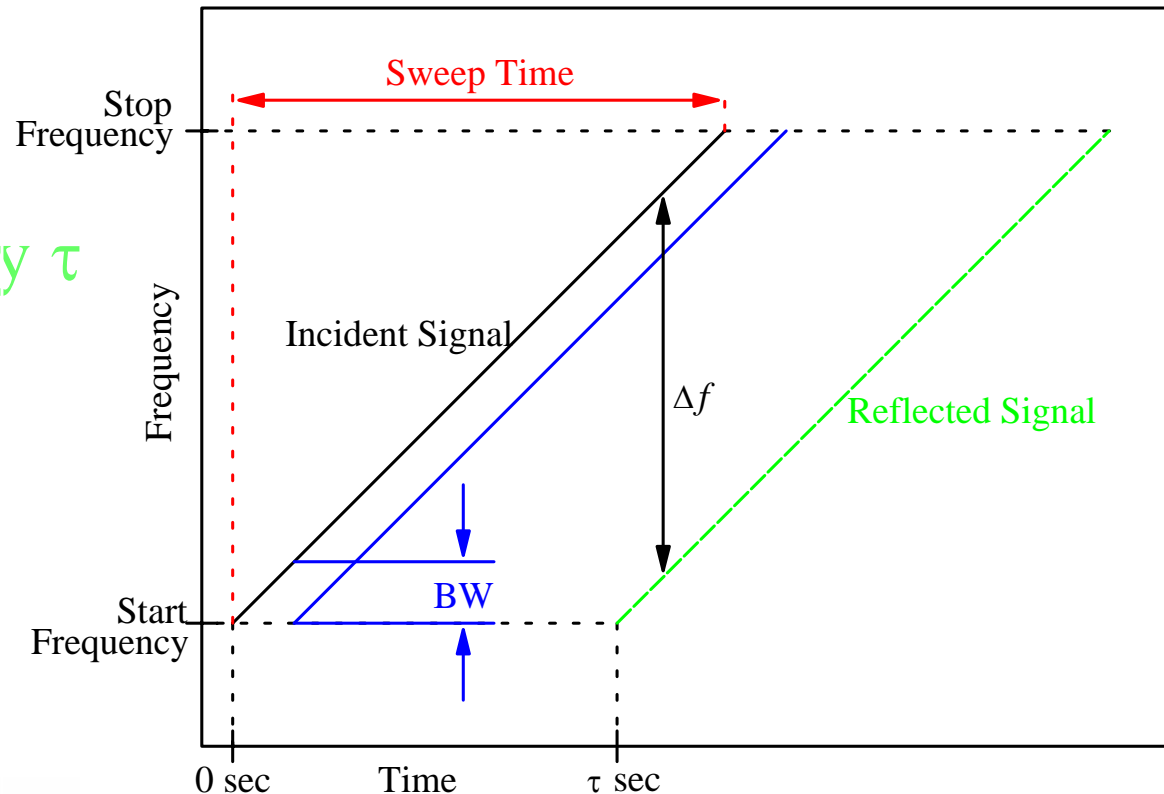
Taking a Measurement: Sweep Time

- In **Sweep Mode** (as opposed to **stepped mode**), the network analyzer transmits a linear chirp in sweep mode.

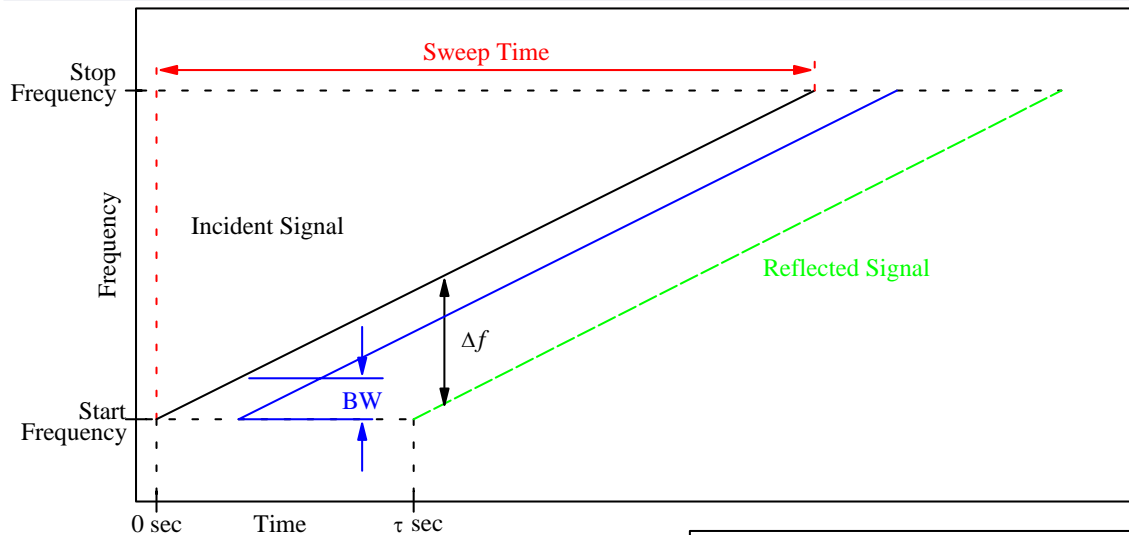


Taking a Measurement: Sweep Time

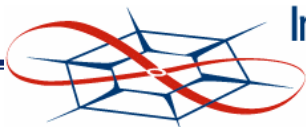
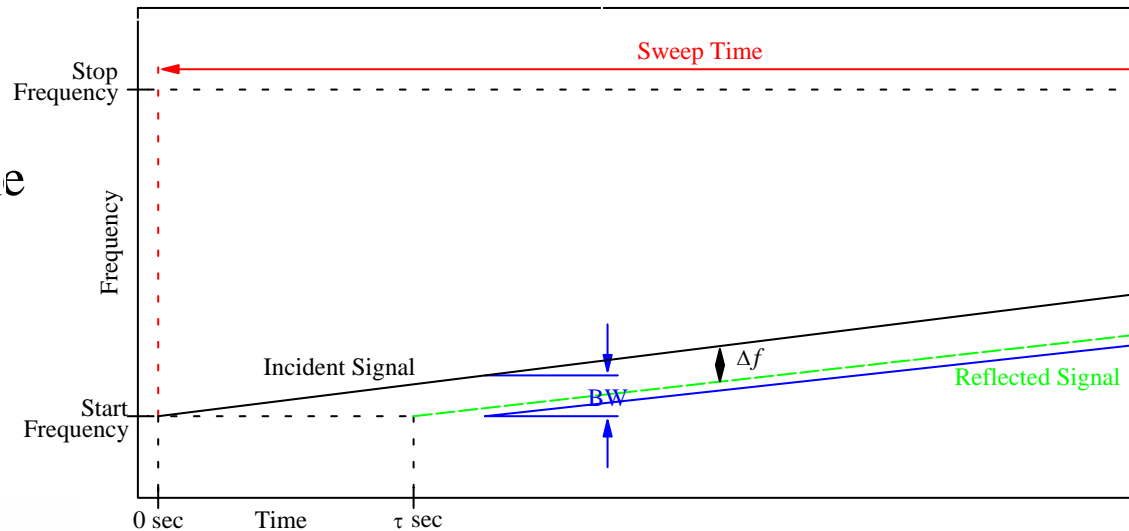
- **Sweep Time:** The time it takes the network analyzer to measure all of its frequencies.
- **IF Bandwidth**
- **Impulse Response Delay τ**



Taking a Measurement: Sweep Time



- **Solution:**
Increase Sweep Time



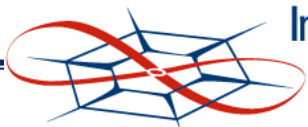
Taking a Measurement: Sweep Time

- Closer Look: We have set our network analyzer to 401 points. At each point, the network analyzer waits 1/300 seconds (that is the typical group delay through a filter with a bandwidth of 300 Hz).

- The impulse response is 511 ns long at most.

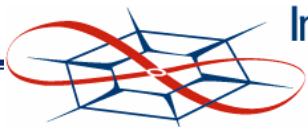
Therefore the sweep rate is:
$$\frac{500 \text{ MHz}}{401 \text{ pts} \cdot \frac{1}{300 \text{ sec}}} = 374 \frac{\text{MHz}}{\text{sec}}$$

- The change in frequency during the impulse response is:
$$374 \frac{\text{MHz}}{\text{sec}} \cdot 511 \text{ nsec} = 191 \text{ Hz} < 300 \text{ Hz}$$



Taking a Measurement: Sweep Time

- **The sweep time is always okay if you are properly sampling the frequency domain!**



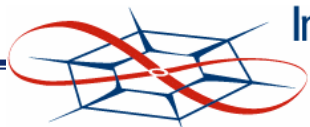
References

- Ballo, David, *Network Analyzer Basics*, Back-to-Basics Seminar, Hewlett-Packard Company, 1997.
- HP 8753D User Guide, Hewlett-Packard Company.
- HP 8720 User Guide, Hewlett-Packard Company.



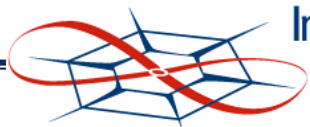
Example

- PRISM SAR example: We want to measure our antennas.
 - There is a total of 100 meters of cable, all with a velocity factor of 75%.
 - The separation between the two antennas is 30 meters and the longest multipath expected is 80 meters.
 - The frequencies of interest are 50-500 MHz.
 - The antennas will receive noise power from radio stations (GIVE LEVEL HERE).
 - The 1 dB compression point of the transmit amplifier is 25 dBm and it has 35 dB of gain.



Example

- Determine start and stop frequency
- Determine transmit power
- Determine number of points
- Determine IF bandwidth and averaging
- Calibrate network analyzer
- Save calibration
- Take measurements



DUT Example

- Filters
 - The local Sunflower cable network uses approximately 1 GHz of bandwidth on their cable network.
 - Each channel occupies 5.5 MHz of this bandwidth.
 - Sunflower offers several options: basic, unlimited, data, etc.
 - The channels that come with the basic package are grouped into the lower part of the spectrum. Therefore, subscribers to the basic service have a low pass filter placed on the cable into their house.
 - The channels used for data are grouped in the upper part of the spectrum. These subscribers have a high pass filter placed on the cable into their house.



Network Analyzer (Simplified)

