

Review Test 1

- Signals & Systems

+ Time

+ Frequency

+ Fourier Trans

Transforms

+ Bandwidth

+ Signals through systems $|H(f)|^2$

+ Power & Energy

+ Period signals - FS

- Power Spectral Density

$$X(f) \rightarrow \boxed{H(f)} \rightarrow Y(f) \quad S_Y(f) = S_X(f) |H(f)|^2$$

$$P_X = \int_{-\infty}^{\infty} S_X(f) df$$

$$P_N = B_N = \frac{1}{2} \int_0^{\infty} |H(f)|^2 df / |H(0)|^2$$

- Ideal Filter

+ LPF

+ BPF

- Distortionless Transmission $y(t) = K x(t - \tau)$

$$|H(f)| = K$$

- Quadrature rep of signals

$$y(t) = X_I(t) \cos(2\pi f_c t) + X_Q(t) \sin(2\pi f_c t)$$

Amp Modulation

- PSB-SC

$$B_{RF} = 2B_X$$

+ Sync receiver $\rightarrow$ carrier recovery
pilot tone

- OSB-LC (Standard AM)

$$y_{RF}(t) = A_c (1 + k_a x(t)) \cos 2\pi f_c t$$

$$k_a |x(t)|_{\max} = m \quad m \leq 1$$

$$+ B_{RF} = 2 B_x$$

$$+ P_c = A_c^2 / 2 \quad P_s = \frac{A_c^2 m^2 P_x}{2} \quad \begin{matrix} P_x = \text{Power} \\ \text{in } x(t) \end{matrix}$$

$$|x(t)|_{\max} = 1$$

$$+ \mu = \frac{P_s}{P_s + P_c} \quad \text{If } x(t) = \cos(2\pi f_m t)$$

$$\mu = \frac{m^2}{2 + m^2}$$

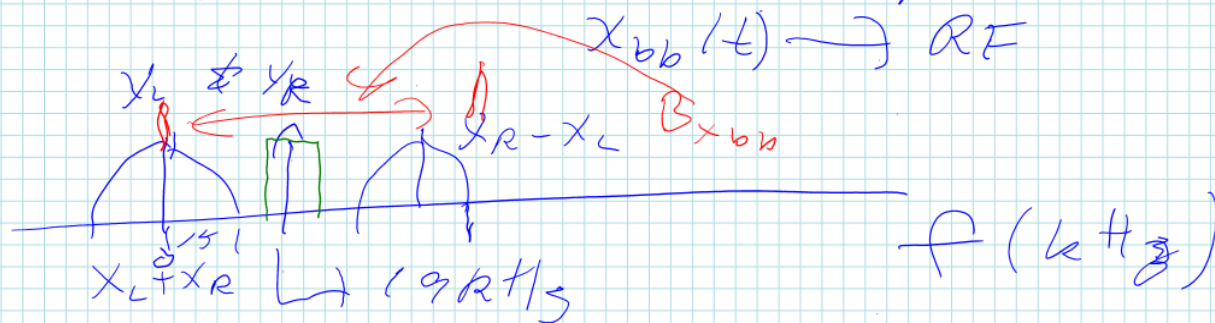
$$\mu_{\max} = \frac{1}{3} \quad m = 1$$

+ Poor Power Efficiency

+ Simple Rec. $\rightarrow$ Envelope Det

+ Poor low frequency response

- FDM $X_i(t) \dots X_n(t) \rightarrow$ use a part of RF spectrum to share
 $+ X_i(t) \rightarrow$ RF f_{sub}

$$+ x_1 \dots x_n(t) = \text{form composite BB signal}$$


Calculate B_{RF} ↓
Sketch RF spectrum

→ Superhot

+ Block Diagram

$$+f_{L0} = f_c^0 + f_{IF}$$

+ BW & center frequency of components

+ Image Frequency

- SSB

$$+ B_{\alpha F} = B_X$$

+ need sync rec

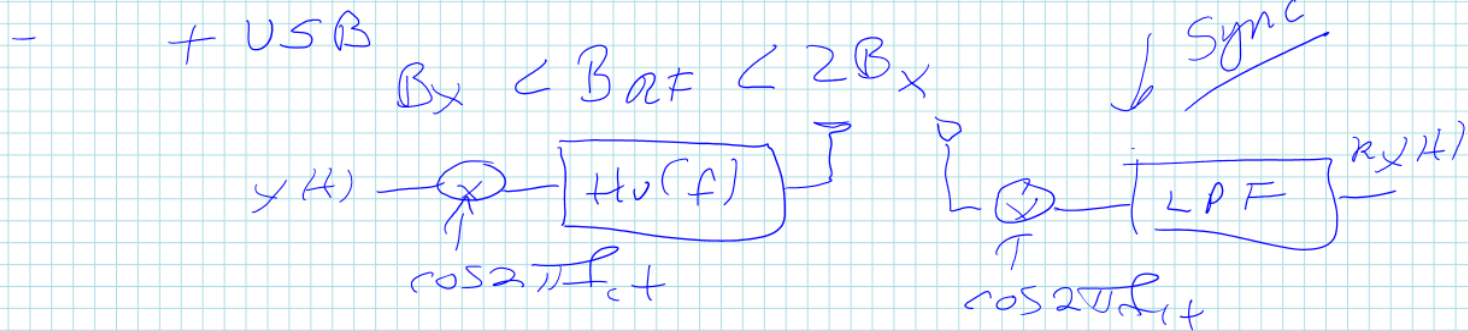
$$+ y_{SSB}(t) = x(t) \cos 2\pi f_c t - \hat{x}(t) \sin(2\pi f_c t)$$

$$y(t) \rightarrow [90^\circ] \rightarrow \hat{x}(t)$$

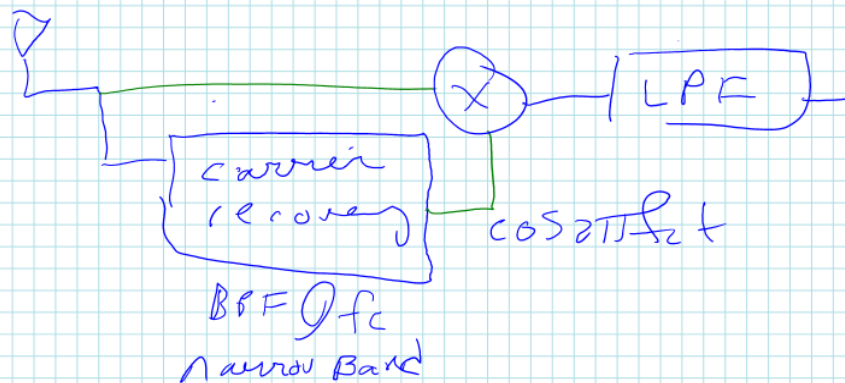
$$+ SS B + L C$$

- * use endcap detector
- ~~the~~ not power efficient

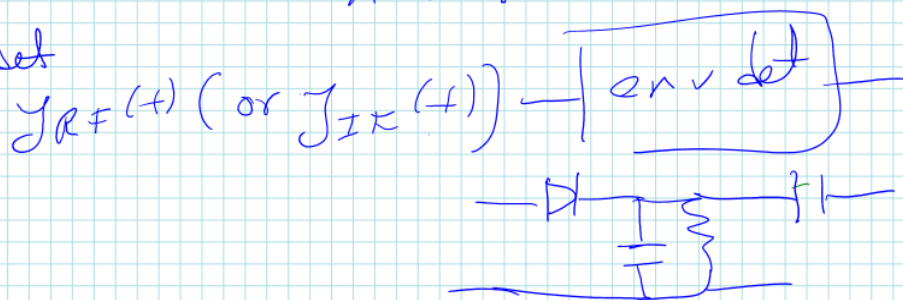
Page 3



+ VSB + LC
Sync rec



Env det



Angle Mod

$$y_{RF}(t) = A_c \cos(\theta(t))$$

$$f_i(t) = \frac{1}{2\pi} \frac{d\theta(t)}{dt} \quad Hz$$

FM $f_i(t) \propto m(t)$

PM $\phi_i(t) \propto m(t)$

FM $y_{FM}(t) = A_c \cos(2\pi f_c t + k_f \int_0^t m(x) dx)$

FM is non linear

$m(t) = A_m \cos(2\pi f_m t)$

$y_{FM}(t) = A_c \cos(2\pi f_c t + \beta \sin(2\pi f_m t))$
 $= A_c \sum_{n=-\infty}^{\infty} J_n(\beta) \cos(2\pi(f_c + n f_m)t)$
 $\beta = \Delta f / f_m \rightarrow$ Table

Parameters $\Delta f = |x(t)|_{\text{peak}} k_f$ for special case

$= A_m k_f$

$\beta = \Delta f / f_m$

Bandwidth

If β small $B_{RF} = 2B_x$

β large $B_{RF} = 2\Delta f$

ROT $B_{RF} = 2B_x(1 + \beta)$

Power $P_T = A_c^2 / 2$

$P_c = (A_c J_0(\beta))^2 / 2$

① $f_c + n f_m$ $P_n = (A_c J_n(\beta))^2 / 2$

$f_c \pm n f_m$ $P_{\text{pair sidebands}} = (A_c J_n(\beta))^2$

Transmitters

+ Indirect $y_{BEM}(t) \rightarrow \text{LP} \rightarrow \text{BPF} \rightarrow$

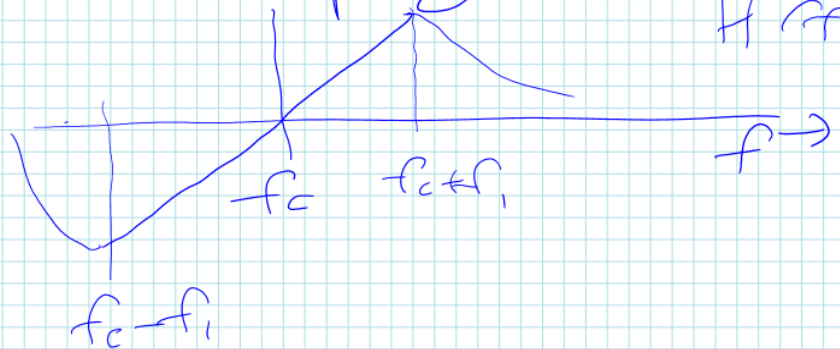
+ Direct $f_i(t) = \frac{v_{in}(t)}{2\pi \sqrt{LC}(t)}$
 $c(t) \propto n(t)$
VCO

Receivers

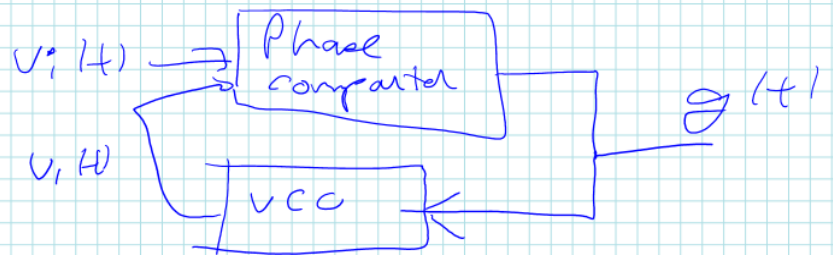
+ $\frac{dy_{em}}{dt} \rightarrow AM$

+ Slope detector

+ Balanced frequency Discriminator H.F.I

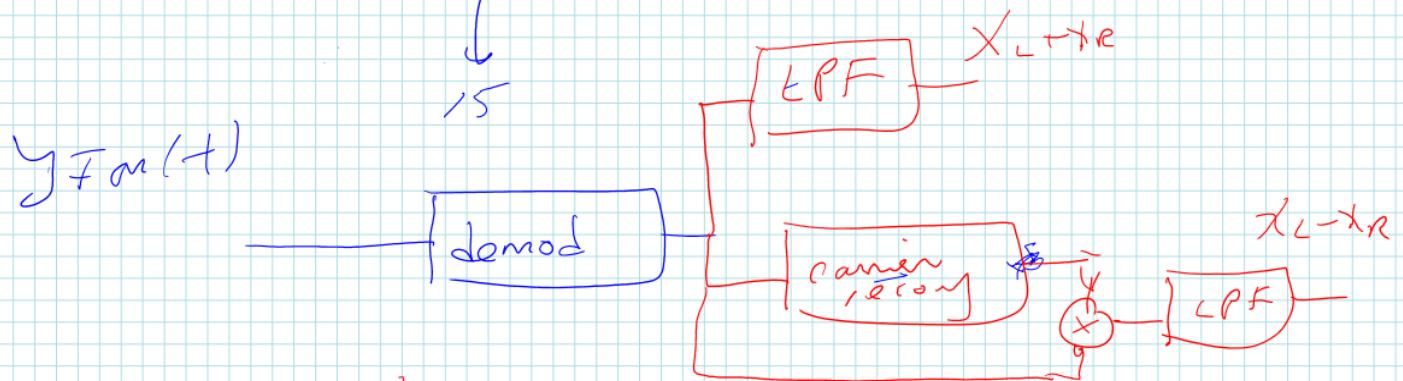
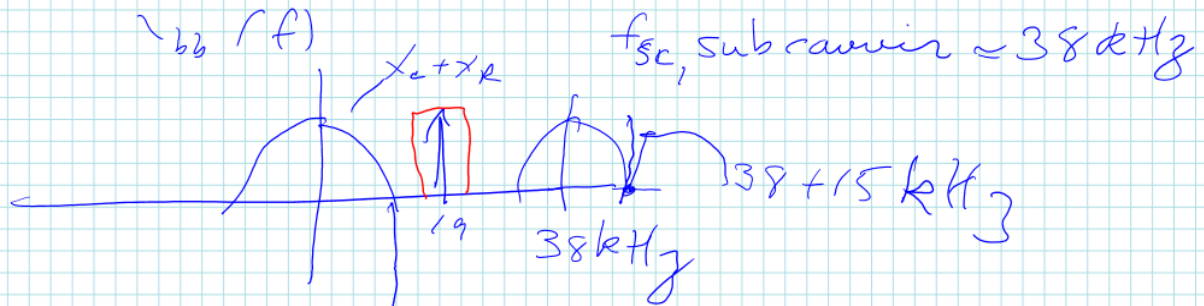
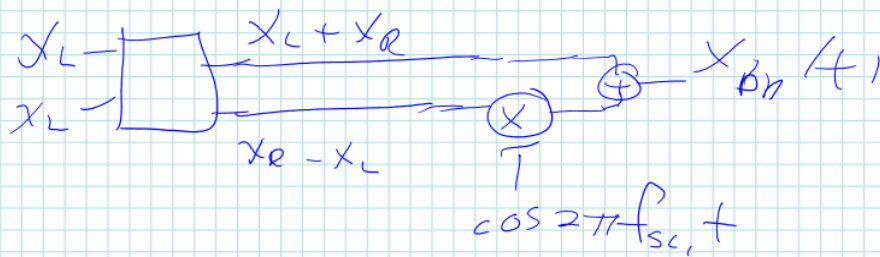


+ PLL



Input $f_i(t)$ of $v_i(t)$
to output $g(t)$

$f_i(t)$ is instantaneous freq of $v_i(t)$



HW #3 problem 1

$$m(t) = \cos(2\pi f_1 t) + 0.5 \cos(2\pi f_2 t) + 0.25 \cos(2\pi f_3 t)$$

