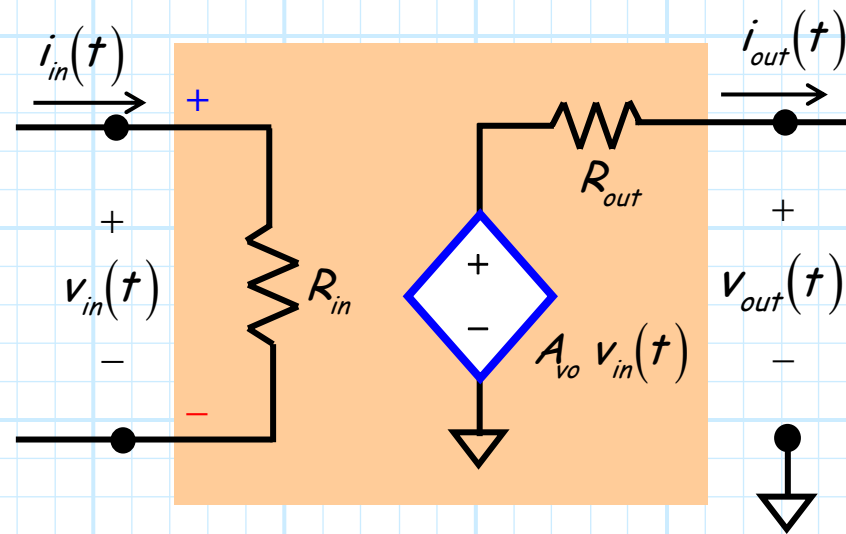


The Ideal Operational Amplifier

We begin by considering the **equivalent circuit** of an **ideal op-amp**:



Note that **output** voltage is defined with respect to **ground** potential, while the input voltage is simply the potential **difference** between the **plus (+)** terminal and the **minus (-)** terminal.

Very large and very small

Of course, we have **three** parameters in this circuit model: input resistance, output resistance, and open circuit voltage gain.

Q: *So what are the **ideal** attributes (R_{in} , R_{out} , and A_{vo}) of an operational amplifier? In other words, what is the **perfect** op-amp?*

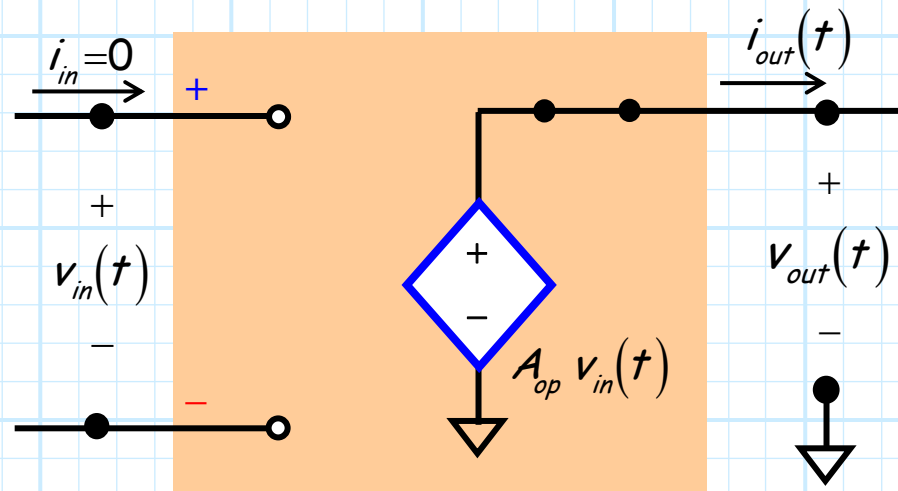
A1: The **input resistance** of a perfect op-amp is infinitely **large** (i.e., $R_{in} = \infty$).

A2: The **output resistance** of a perfect op-amp is **zero** (i.e., $R_{out} = 0$).

A3: The open-circuit **voltage gain** of a perfect op-amp is **very large**, approaching **infinity** ($A_{vo} \cong \infty$)!

The ideal op-amp model

Thus, the equivalent circuit model of an **ideal op-amp** is:



Here we have **changed the notation** of the open-circuit voltage gain.

The value A_{op} is used, where:

$$A_{op} \doteq \lim_{A_{vo} \rightarrow \infty} A_{vo}$$

In other words, the gain value A_{op} is **unfathomably large!**

Ideal at all frequencies!

Note then:

1. Since the input resistance is infinite, the **input current is zero**—always!
2. Since the output resistance is zero, the **output voltage is equal** to the **open-circuit** output voltage, even when the output load is **not** an open circuit! I.E.:

$$v_{out}(t) = A_{op} v_{in}(t) \quad \leftarrow \text{regardless of } i_{out}!$$

Q: *What about the **bandwidth** of this "ideal" op-amp; is the model **only** valid for **low-frequencies**?*

A: Not for an **ideal** op-amp!

The bandwidth of an **ideal** op-amp is likewise **infinite**.

It just seems so perfect

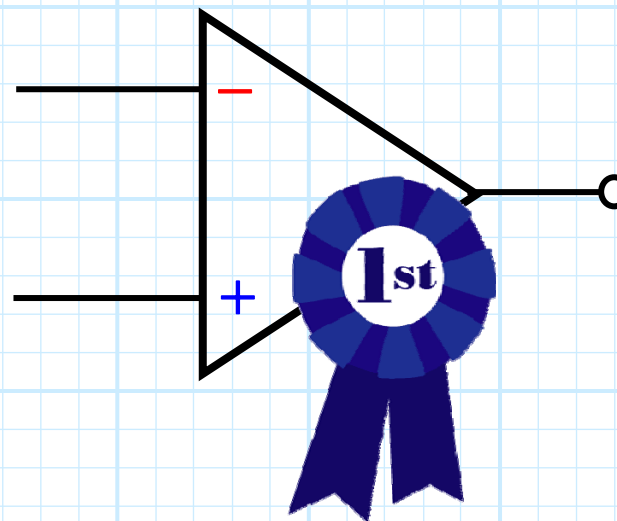
Q: Wow! Unfathomably **high** voltage gain, **infinite** input resistance (impedance), **zero** output resistance (impedance), and:

$$v_{out}(t) = A_{op} v_{in}(t)$$

regardless of the frequency spectrum $V_{in}(\omega)$.

*This sounds like the **perfect voltage amplifier**!*

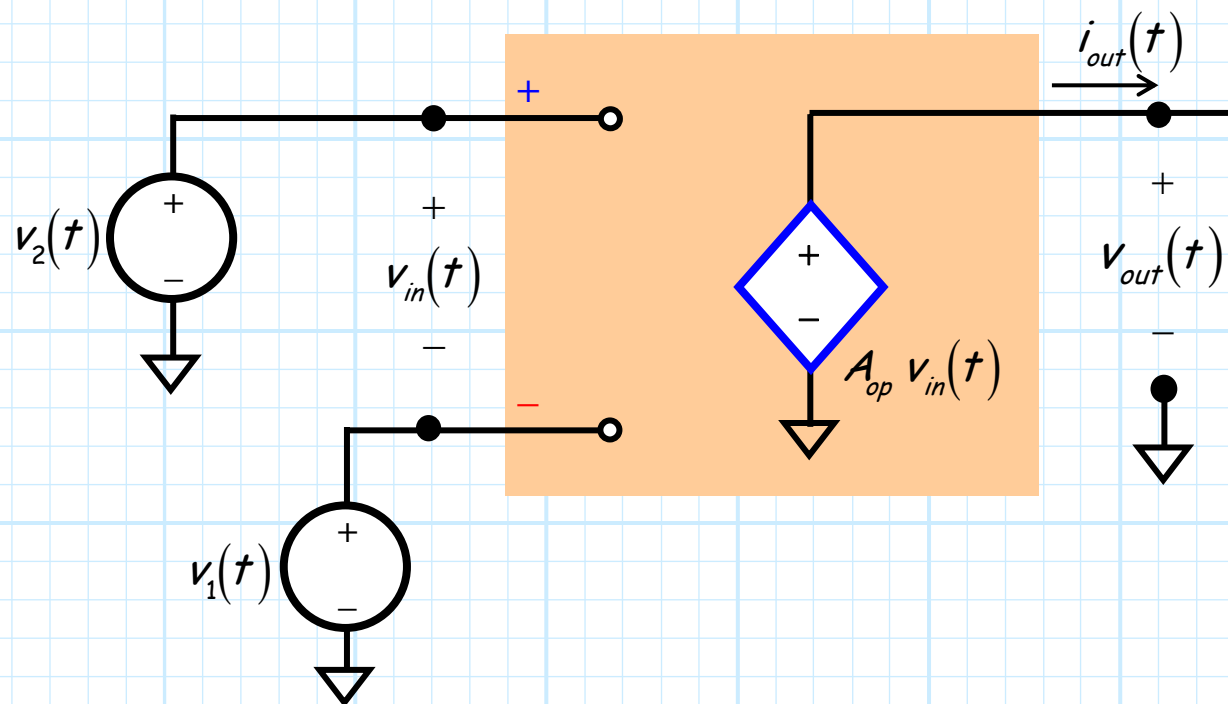
A: It is! That's why we refer to it as the **ideal** op-amp.



Why the output but not the input?

Q: *So why isn't the input voltage with respect to ground potential? Why is not the minus (-) input terminal **connected to ground**?*

A: Generally speaking, we find that **two different voltages** will be connected to the two different input terminals:



The input is a differential voltage

From KVL it is clear (right?) that the **input voltage** is:

$$v_{in}(t) = v_2(t) - v_1(t)$$

And so the **output voltage** is:

$$v_{out}(t) = A_{op}(v_2(t) - v_1(t))$$

Note that the input voltage is simply the **difference** between the **two** input signals v_2 and v_1 .

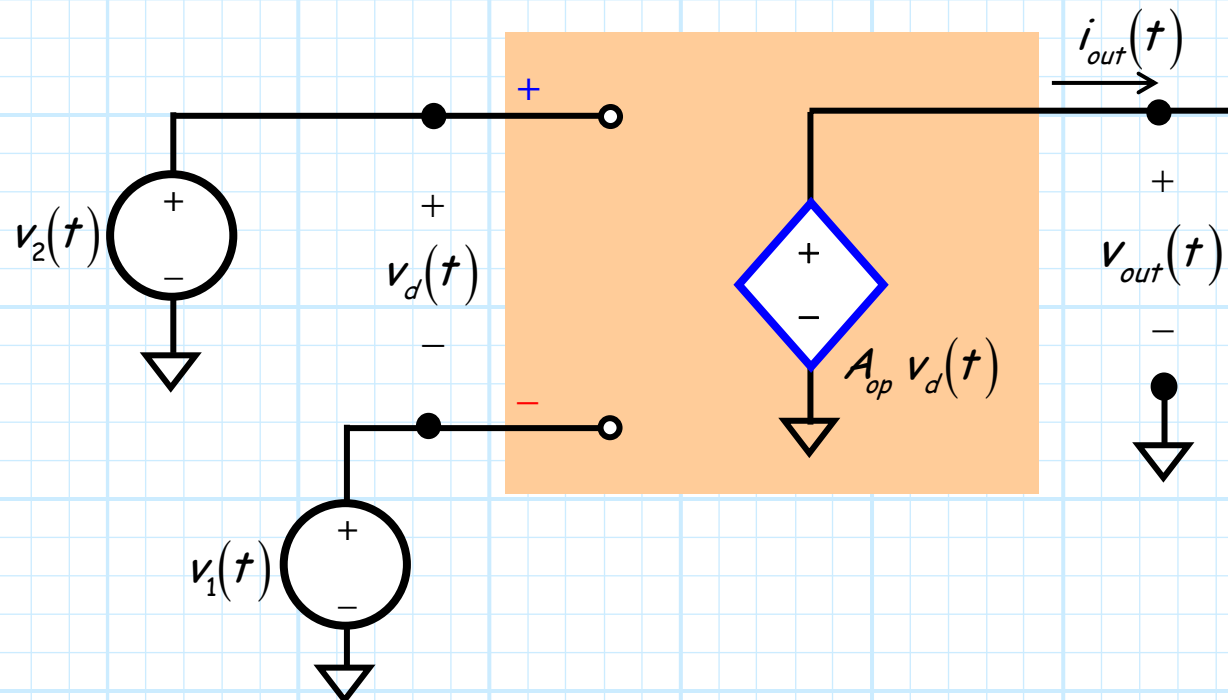
We call this the **differential** input signal:

$$v_d(t) \doteq v_2(t) - v_1(t)$$

It's called a differential amplifier

Thus, we can likewise express the output as:

$$v_{out}(t) = A_{op} v_d(t)$$



Amplifiers of this type—where the input voltage is **not** defined with respect to ground—are referred to as **differential amplifiers**, as they can amplify the **differential mode** of two distinct signals (e.g., $v_2(t)$ and $v_1(t)$).

An example

For **example**, say:

$$v_1(t) = 7.0 \cos(10\pi t) + 2.0 \cos(5\pi t)$$

and:

$$v_2(t) = 7.0 \cos(10\pi t) + 5.0 \cos(5\pi t)$$

the **ideal** op-amp output voltage is therefore:

$$\begin{aligned} v_{out}(t) &= A_{op} (v_2(t) - v_1(t)) \\ &= A_{op} 3.0 \cos(5\pi t) \end{aligned}$$

➔ What happened to $\cos(10\pi t)$??

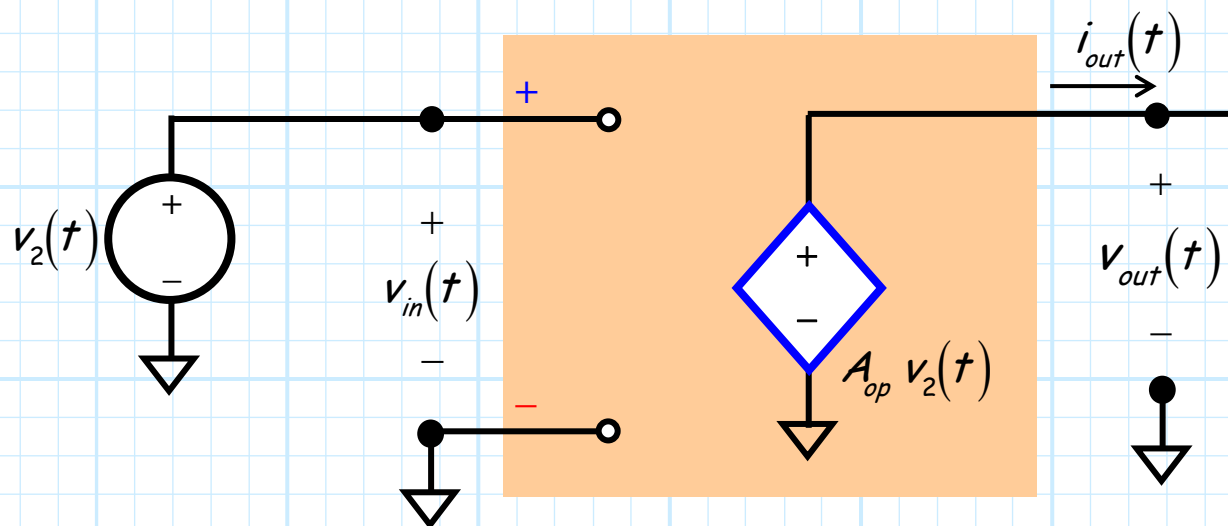
The common mode disappears!

Note:

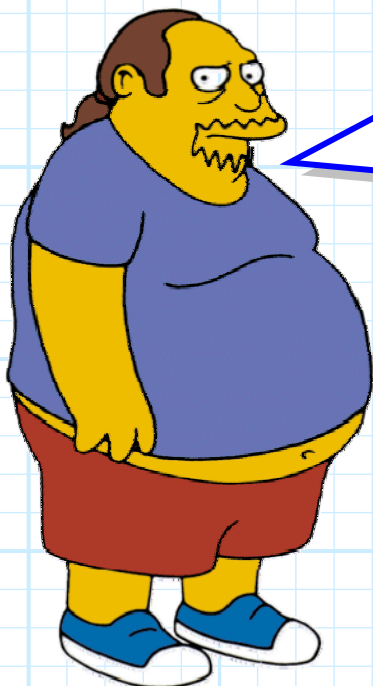
1. The **difference** between $v_2(t)$ and $v_1(t)$ is amplified.
2. The **common** signal ($7 \cos 10\pi t$) is eliminated by the subtraction.

☞ Difference amplifiers **ideally** have perfect **common-mode rejection**. That is, the common signal between the two inputs has **no effect** on the output signal.

Of course, we can always **connect** a terminal to ground potential (i.e., $v_1(t) = 0$), thus making the input voltage a value with respect to ground:



The ideal op-amp; is it bogus?



Q: *I scoff at your so-called "ideal" op-amp.*

*Although $R_{in} = \infty$ and $R_{out} = 0$ are obviously correct, I deem your assertion that A_{op} should be **unfathomably large** (approaching ∞) to be a **silly** notion.*

*After all, a gigantic gain A_{op} would mean that the **output voltage** $v_{out} = A_{op}(v_2 - v_1)$ would likewise be **unfathomably large**—the **destructive** implications are **obvious**.*

A: It is true that the output voltage will be very large—**unless** the differential voltage is **unfathomably small**!

Definitely not bogus!

For example, what if the differential voltage is approximately (i.e., almost) **zero**:

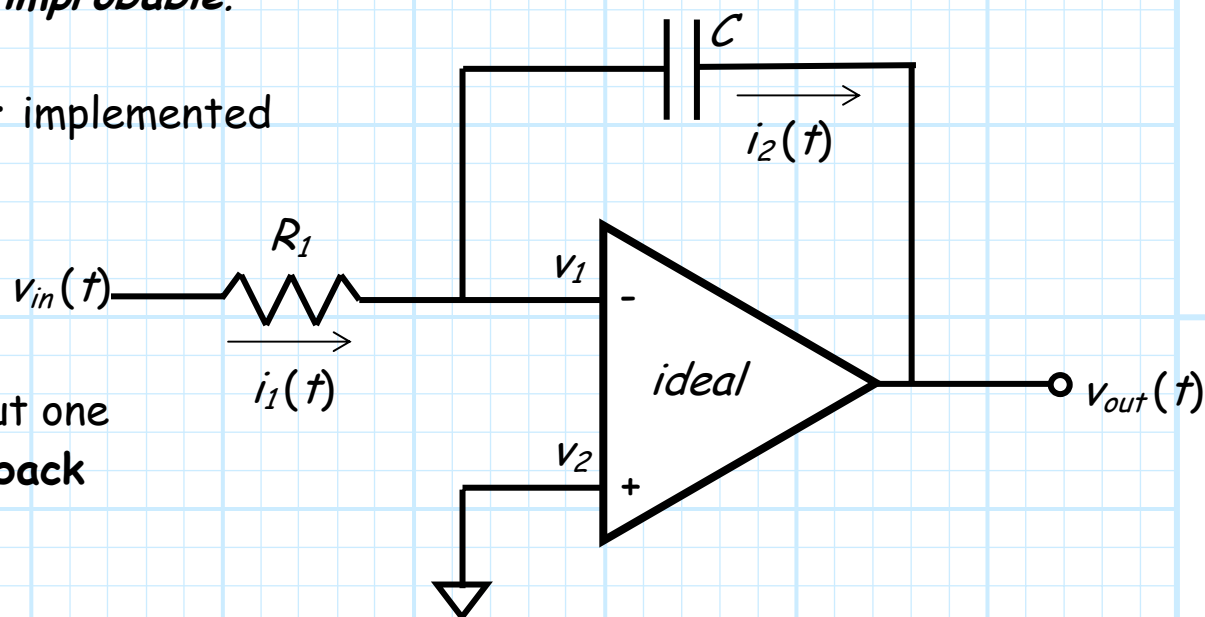
$$v_d(t) \cong 0 \quad \Rightarrow \quad v_2(t) \cong v_1(t) \quad ?$$

In this case, the output voltage may **not** be very large at all!

Q: Yes, but what is the *likelihood* that the two voltages $v_2(t)$ and $v_1(t)$ are nearly the same? This seems *improbable*.

A: Op-amps are generally **not** implemented by themselves!

Instead, they typically are but one component of **many** in a **feedback amplifier**.



Get used to the virtual short!

In these applications, we will **indeed** find that $v_2 \cong v_1$ —but we will also find that this is a **desirable** condition!

The condition $v_2 \cong v_1$ is known as a **virtual short**.

If this is not true, the output **voltage** of an ideal op-amp will be **unfathomably large**.



As a result, the **virtual short** $v_2 \cong v_1$ is almost **always** the case in useful op-amp circuits.