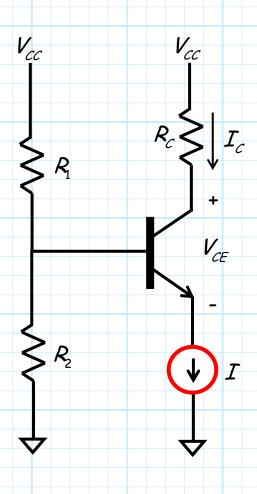
BJT Biasing using a Current Source

Another way to bias a BJT small signal amplifier is to use one voltage source and one current source. This biasing scheme has a number of important advantages:



1. The DC emitter current is independent of β or BJT temperature!

Therefore, the DC collector current $I_{\mathcal{C}} = \alpha I_{\mathcal{E}} \approx I_{\mathcal{E}}$ is nearly independent of these parameters as well.

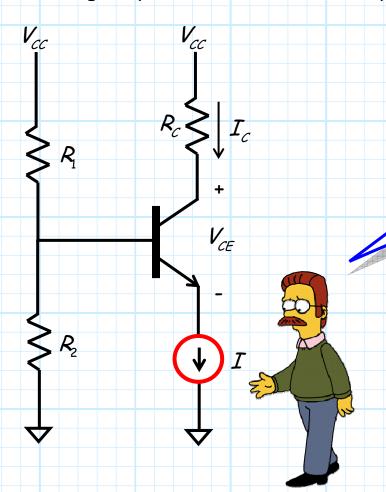
2. This means that the emitter voltage can be set at an arbitrarily low value!

Therefore, the output voltage swing can be much larger than an equivalent single-supply amplifier!

3. We can make resistors R_1 and R_2 large without making design sensitive to temperature and β .

The current source: not as easy as it appears

Note that ideally, we would set the emitter votage to zero ($V_{\varepsilon}=0$), and thus the collector voltage to $V_c=V_{cc}/2$ to maximize the output swing (i.e., maximize the largest possible undistorted output signal).



Q: But, isn't it diddly darn difficult to actually build an ideal current source!?

A: True! For reasons we shall study later, most current sources require a minimum voltage across them in order to operate properly.

Put collector voltage half way between floor and ceiling

Thus, our bias rule should be:

Make the DC emitter voltage V_{ε} as **small** as possible (and still have the current source work!).

Then set the **current source** to a value equal to the desired DC collector current (i.e., $I_C \approx I_F$):

$$I = I_{\mathcal{E}} \approx I_{\mathcal{C}}$$

To maximize the output voltage swing, we still want to place the DC collector voltage V_c half way between V_{cc} and V_E .

$$V_C = \frac{V_{CC} + V_E}{2}$$

The collector resistor therefore should be:

$$R_{c} = \frac{V_{cc} - V_{c}}{I_{c}} = \frac{V_{cc} - V_{c}}{I} = \frac{V_{cc} - V_{E}}{2I}$$

R1 and R2: same as before

The remaining resistors R_1 and R_2 are determined in the same manner as with the single-supply bias design, i.e.:

$$R_1 = \frac{V_{CC} - V_B}{I_1}$$

and

$$R_2 = rac{V_B}{I_2} pprox rac{V_B}{I_1}$$

where the base voltage is approximately:

$$V_{B}=0.7+V_{E}$$

and the current I_1 is any value in the range:

$$0.1\,I_{\mathcal{C}} < I_1 < I_{\mathcal{C}}$$

Just the kind of subtle topic I might put on an exam

For example, say we wish to design a biasing network where:

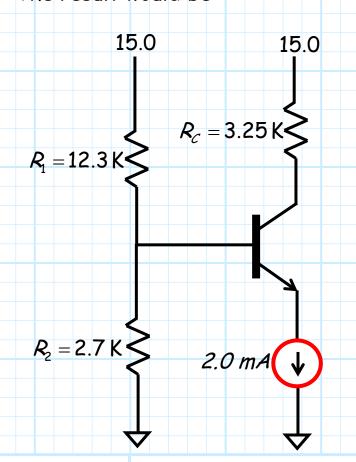
$$I_c = 2 mA$$

$$V_{\mathcal{E}} \geq 2.0 V$$

$$I_c = 2 \, mA$$
 $V_E \ge 2.0 \, V$ $V_{cc} = 15.0 \, V$ $I_1 = 0.5 \, I_c$

$$I_1 = 0.5 I_C$$

The result would be:



It is obvious to me that this bias design satisfies the parameters described above.

But, don't take my word for it verify for yourself that these resistor values are correct.