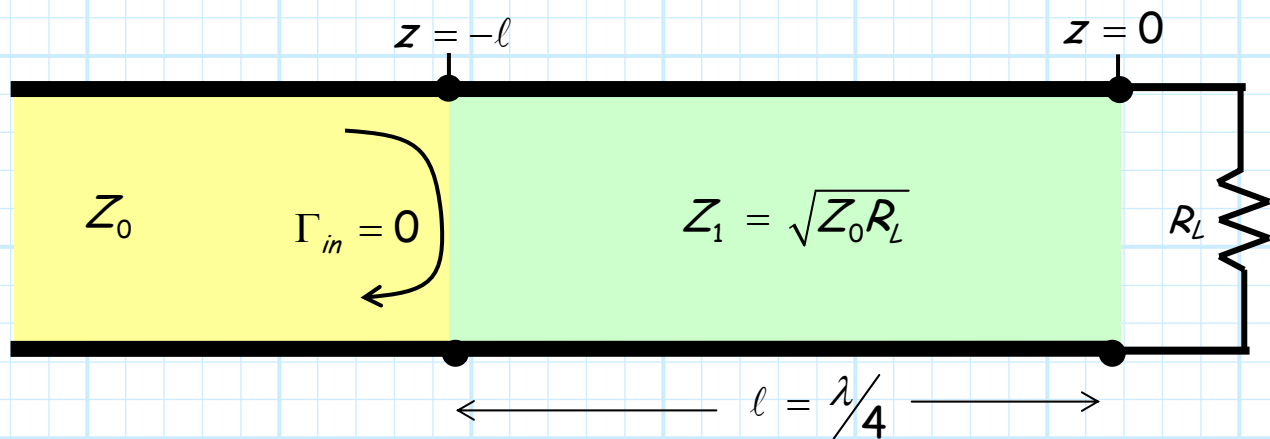


# Multiple Reflection Viewpoint

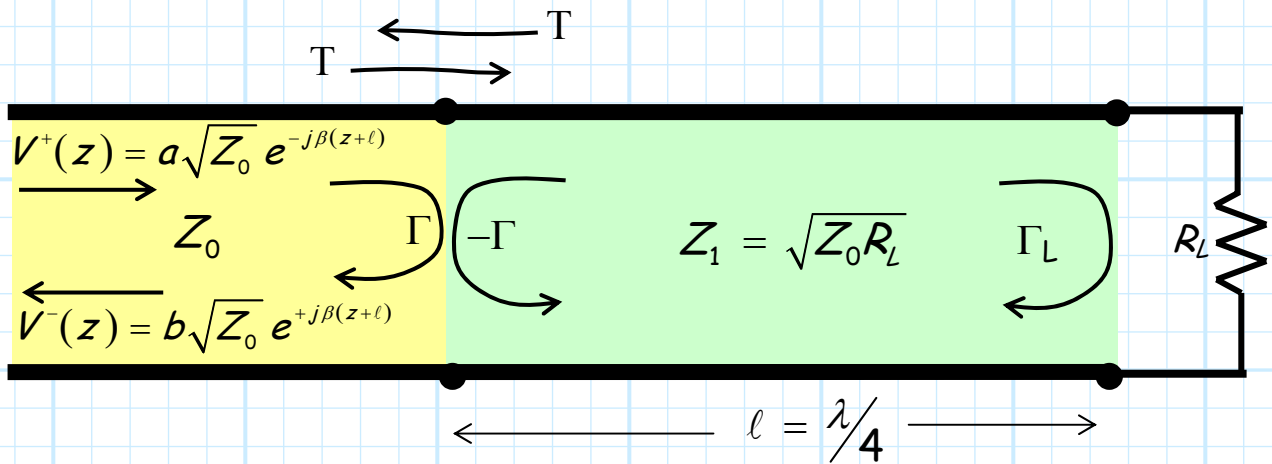
The **quarter-wave** transformer brings up an interesting question in  $\mu$ -wave engineering.



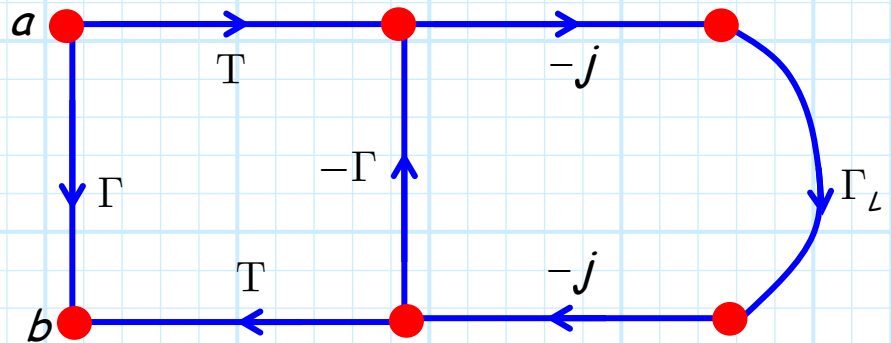
**Q:** *Why is there **no** reflection at  $z = -l$  ? It appears that the line is **mismatched** at both  $z = 0$  and  $z = -l$ .*

**A:** In fact there **are** reflections at these mismatched interfaces—an **infinite** number of them!

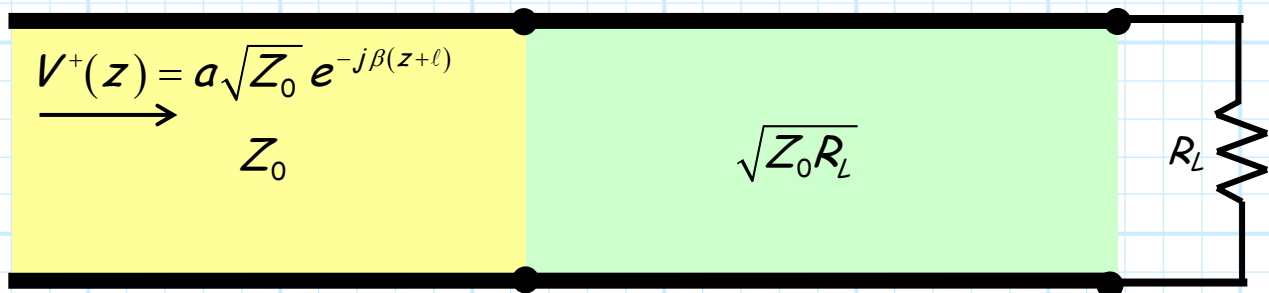
We can use our **signal flow graph** to determine the propagation series, once we determine all the **propagation paths** through the quarter-wave transformer.



$$b = a \sum_{n=1}^{\infty} p_n$$

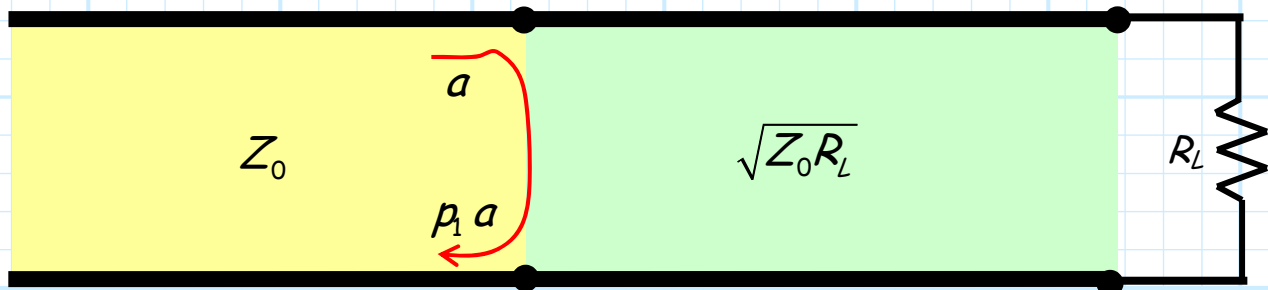


Now let's try to interpret what **physically** happens when the **incident** voltage wave:

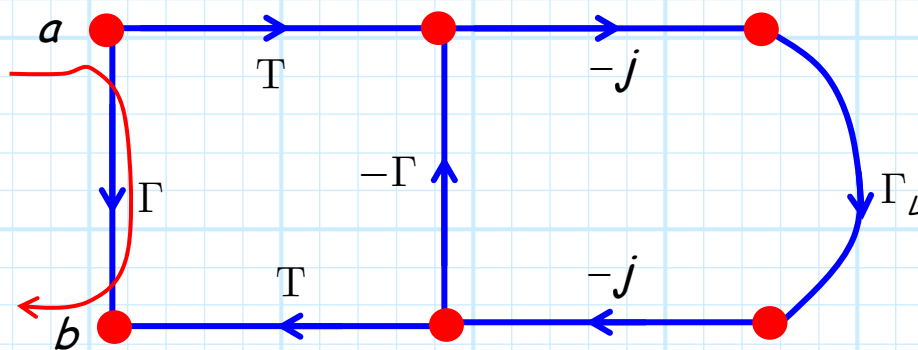


reaches the interface at  $z = -\ell$ . We find that there are **two forward paths** through the quarter-wave transformer signal flow graph.

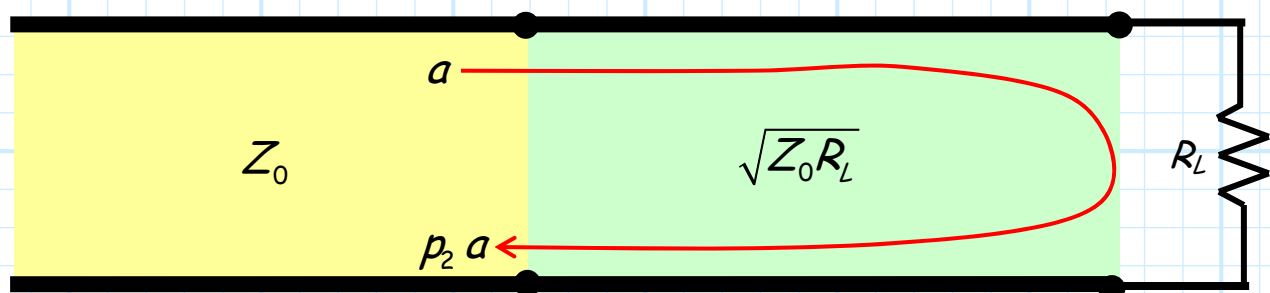
**Path 1.** At  $z = -l$ , the characteristic impedance of the transmission line changes from  $Z_0$  to  $Z_1$ . This mismatch creates a **reflected** wave, with complex amplitude  $\rho_1 a$ :



So,  $\rho_1 = \Gamma$ .



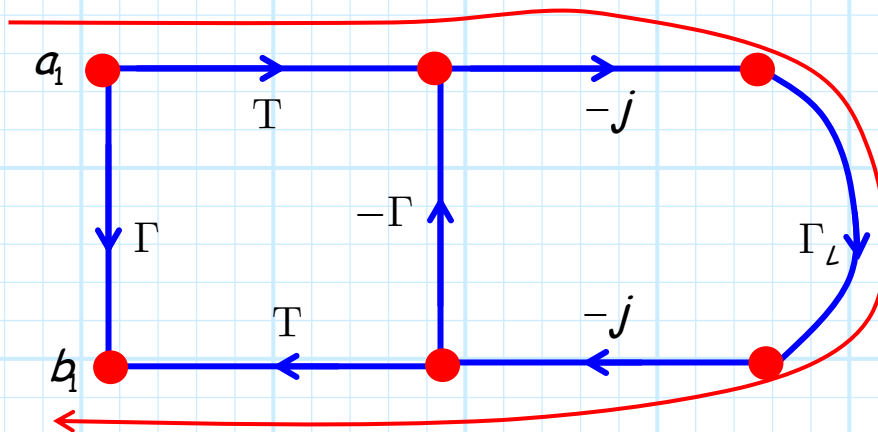
**Path 2.** However, a **portion** of the incident wave is transmitted (T) across the interface at  $z = -l$ , this wave travels a distance of  $\beta l = 90^\circ$  to the load at  $z = 0$ , where a portion of it is reflected ( $\Gamma_L$ ). This wave travels back  $\beta l = 90^\circ$  to the interface at  $z = -l$ , where a portion is again transmitted (T) across into the  $Z_0$  transmission line—**another** reflected wave!



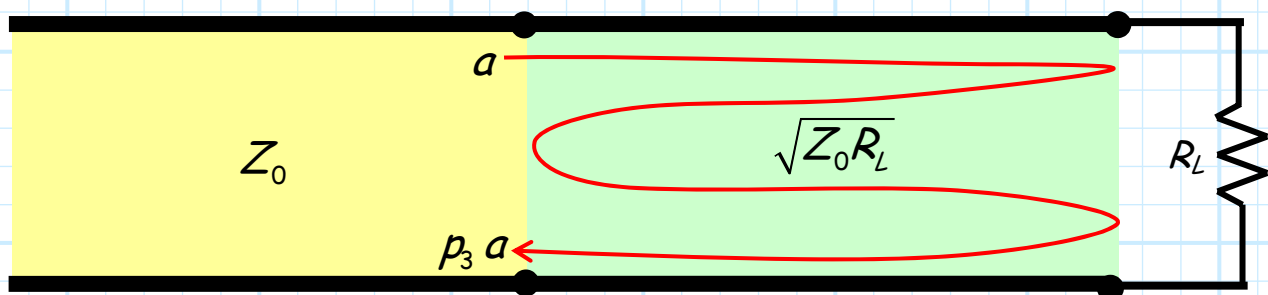
So the **second direct path** is

$$p_2 = T e^{-j90^\circ} \Gamma_L e^{-j90^\circ} T = -T^2 \Gamma_L$$

note that traveling  $2\beta l = 180^\circ$  has produced a **minus** sign in the result.



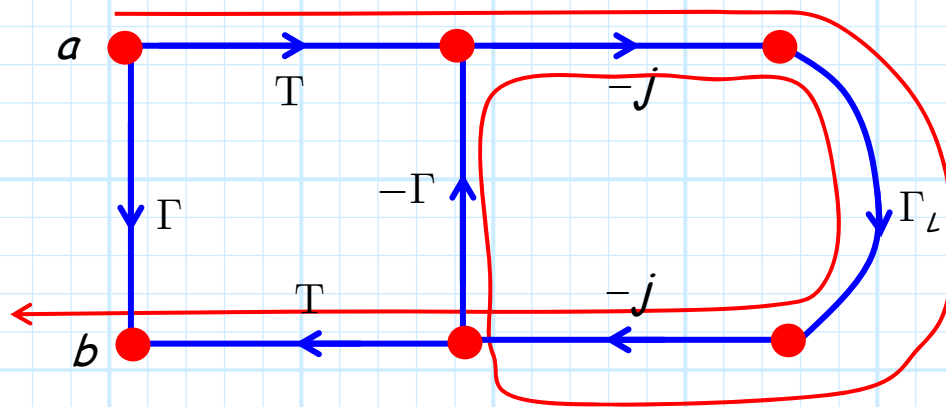
**Path 3.** However, a **portion** of this **second** wave is also **reflected** ( $\Gamma$ ) back into the  $Z_1$  transmission line at  $z = -l$ , where it again travels to  $\beta l = 90^\circ$  the load, is partially reflected ( $\Gamma_L$ ), travels  $\beta l = 90^\circ$  back to  $z = -l$ , and is partially transmitted into  $Z_0$  ( $T$ )—our **third** reflected wave!



where:

$$p_3 = T e^{-j90^\circ} \Gamma_L e^{-j90^\circ} (-\Gamma) e^{-j90^\circ} \Gamma_L e^{-j90^\circ} T$$

$$= -T^2 (\Gamma_L)^2 \Gamma$$



Note that path 3 is **not** a direct path!

**Path  $n$ .** We can see that this “bouncing” back and forth can go on **forever**, with each trip launching a **new** reflected wave into the  $Z_0$  transmission line.

Note however, that the **power** associated with each successive reflected wave is **smaller** than the previous, and so eventually, the power associated with the reflected waves will **diminish** to insignificance!

**Q:** *But, why then is  $\Gamma = 0$  ?*

**A:** Each reflected wave is a **coherent** wave. That is, they all oscillate at same frequency  $\omega$ ; the reflected waves differ only in terms of their **magnitude** and **phase**.

Therefore, to determine the **total** reflected wave, we must perform a **coherent summation** of each reflected wave—this summation of course results in our **propagation series**, a series that must converge for passive devices.

$$b = a \sum_{n=1}^{\infty} p_n$$

It can be shown that the infinite propagation series for **this** quarter-wavelength structure **converges** to the closed-form expression:

$$\frac{b}{a} = \sum_{n=1}^{\infty} p_n = \frac{\Gamma - \Gamma^2 \Gamma_L - \Gamma^4 \Gamma_L}{1 - \Gamma^2}$$

Thus, the **input** reflection coefficient is:

$$\Gamma_{in} = \frac{b}{a} = \frac{\Gamma - \Gamma^2 \Gamma_L - \Gamma^4 \Gamma_L}{1 - \Gamma^2}$$

Using our definitions, it can likewise be shown that the **numerator** of the above expression is:

$$\Gamma - \Gamma^2 \Gamma_L - \Gamma^4 \Gamma_L = \frac{2(Z_1^2 - Z_0 R_L)}{(Z_1 + Z_0)(R_L + Z_1)}$$

It is evident that the numerator (and therefore  $\Gamma$ ) will be **zero** if:

$$Z_1^2 - Z_0 R_L = 0 \quad \Rightarrow \quad Z_1 = \sqrt{Z_0 R_L}$$

**Just** as we expected!

Physically, this results insures that all the reflected waves add coherently together to produce a **zero value!**

Note **all** of our transmission line analysis has been **steady-state** analysis. We assume our signals are **sinusoidal**, of the form  $\exp(j\omega t)$ . Note this signal exists for **all time**  $t$ —the signal is

assumed to have been "on" **forever**, and assumed to continue on forever.

In other words, in steady-state analysis, **all** the multiple reflections have long since occurred, and thus have reached a steady state—the reflected wave is **zero!**