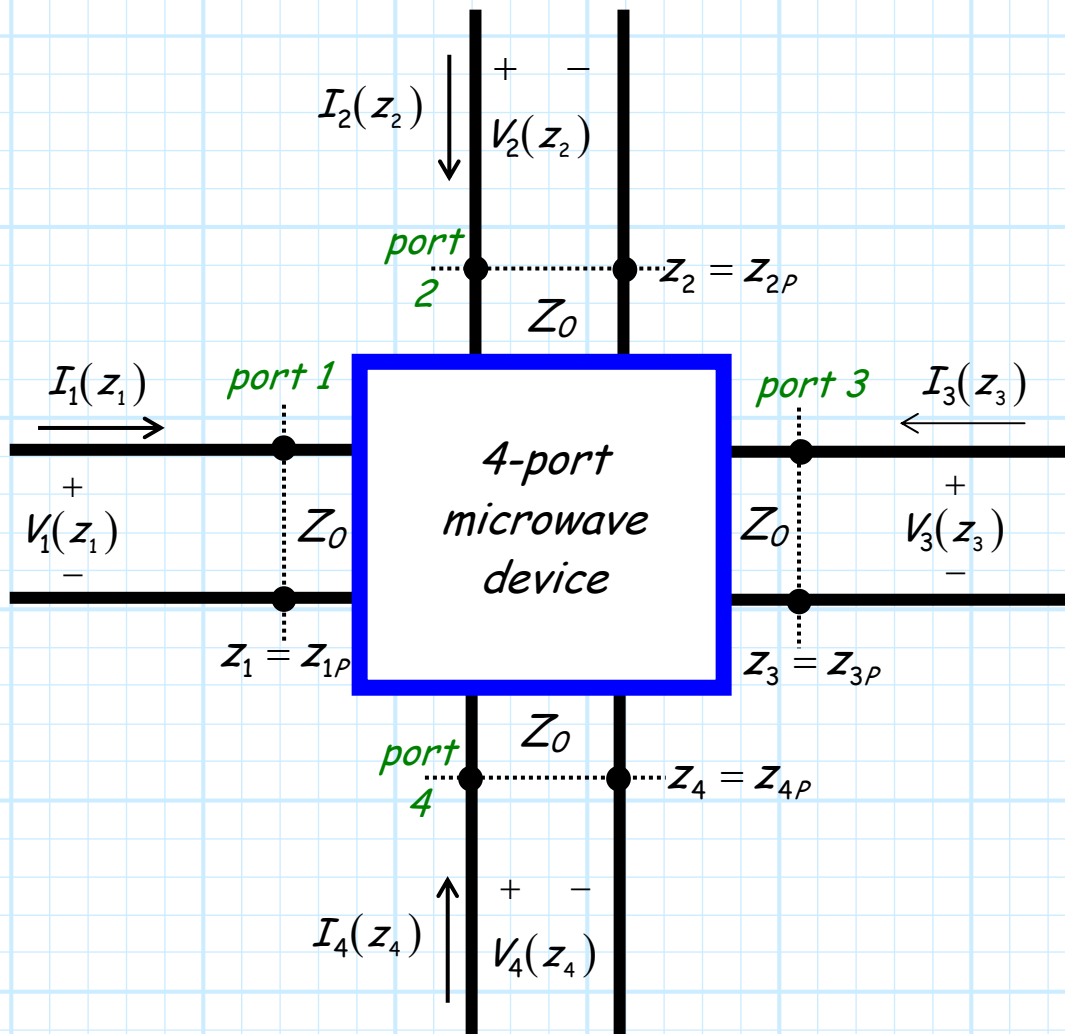


The Admittance Matrix

Consider again the 4-port microwave device shown below:

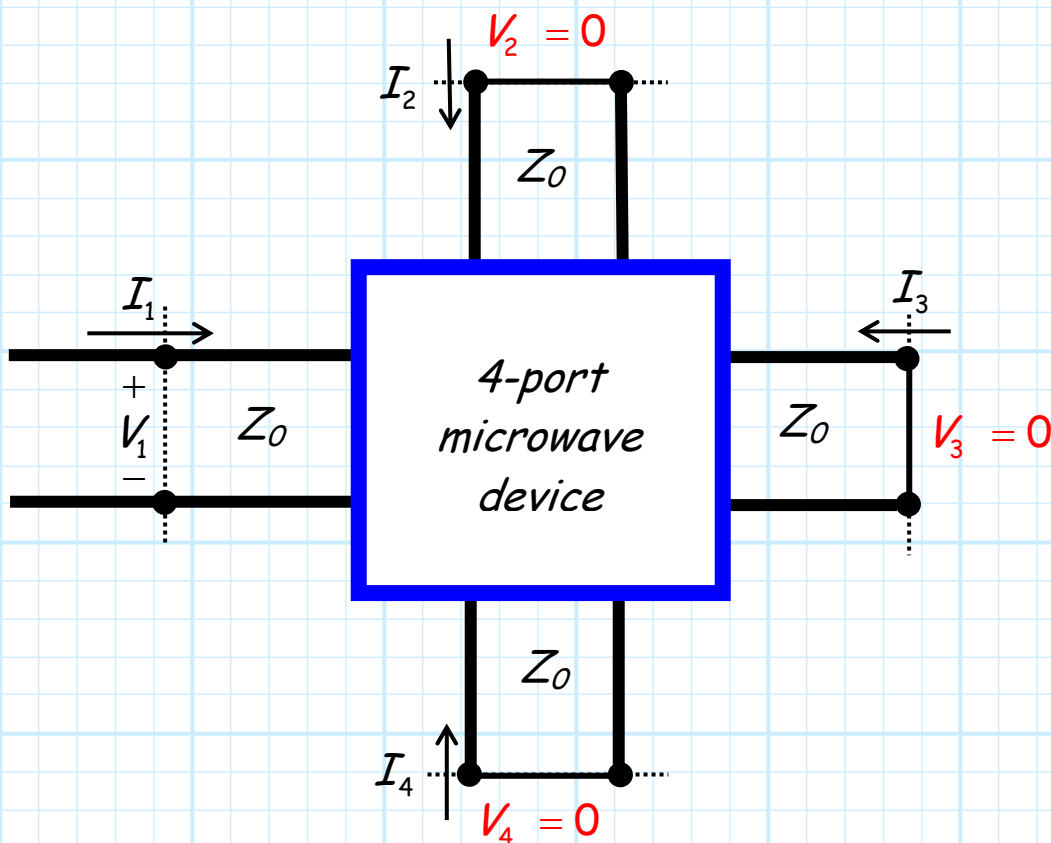


In addition to the Impedance Matrix, we can fully characterize this linear device using the **Admittance Matrix**.

The elements of the Admittance Matrix are the **trans-admittance** parameters Y_{mn} , defined as:

$$y_{mn} = \frac{I_m}{V_n} \quad (\text{given that } V_k = 0 \text{ for all } k \neq n)$$

Note here that the **voltage** at all but one port **must** be equal to **zero**. We can ensure that by simply placing a **short circuit** at these zero voltage ports!



Note that $y_{mn} \neq 1/Z_{mn}$!

Now, we can thus **equivalently** state the definition of trans-admittance as:

$$Y_{mn} = \frac{I_m}{V_n} \quad (\text{given that all ports } k \neq n \text{ are short-circuited})$$

Just as with the trans-impedance values, we can use the trans-admittance values to evaluate general circuit problems, where **none** of the ports have zero voltage.

Since the device is **linear**, the current at any **one** port due to **all** the port currents is simply the coherent **sum** of the currents at that port due to **each** of the port voltages!

For example, the **current** at port 3 can be determined by:

$$I_3 = Y_{34} V_4 + Y_{33} V_3 + Y_{32} V_2 + Y_{31} V_1$$

More **generally**, the current at port m of an N -port device is:

$$I_m = \sum_{n=1}^N y_{mn} V_n$$

This expression can be written in **matrix** form as:

$$\mathbf{I} = \mathcal{Y} \mathbf{V}$$

Where \mathbf{I} is the **vector**:

$$\mathbf{I} = [I_1, I_2, I_3, \dots, I_N]^T$$

and \mathbf{V} is the **vector**:

$$\mathbf{V} = [V_1, V_2, V_3, \dots, V_N]^T$$

And the matrix \mathcal{Y} is called the **admittance matrix**:

$$\mathcal{Y} = \begin{bmatrix} Y_{11} & \dots & Y_{1n} \\ \vdots & \ddots & \vdots \\ Y_{m1} & \dots & Y_{mn} \end{bmatrix}$$

The admittance matrix is a N by N matrix that **completely characterizes** a linear, N -port device. Effectively, the admittance matrix describes a multi-port device the way that Y_L describes a single-port device (e.g., a load)!



But **beware!** The values of the admittance matrix for a particular device or network, just like Y_L , are **frequency dependent!** Thus, it may be more instructive to **explicitly** write:

$$\mathcal{Y}(\omega) = \begin{bmatrix} Y_{11}(\omega) & \dots & Y_{1n}(\omega) \\ \vdots & \ddots & \vdots \\ Y_{m1}(\omega) & \dots & Y_{mn}(\omega) \end{bmatrix}$$

Q: You said earlier that $Y_{mn} \neq 1/Z_{mn}$. Is there any **relationship** between the admittance and impedance matrix of a given device?

A: I don't know! Let's see if **we** can figure it out.

Recall that we can determine the inverse of a matrix. Denoting the matrix inverse of the admittance matrix as \mathcal{Y}^{-1} , we find:

$$\begin{aligned}\mathbf{I} &= \mathcal{Y} \mathbf{V} \\ \mathcal{Y}^{-1} \mathbf{I} &= \mathcal{Y}^{-1} (\mathcal{Y} \mathbf{V}) \\ \mathcal{Y}^{-1} \mathbf{I} &= (\mathcal{Y}^{-1} \mathcal{Y}) \mathbf{V} \\ \mathcal{Y}^{-1} \mathbf{I} &= \mathbf{V}\end{aligned}$$

Meaning that:

$$\mathbf{V} = \mathcal{Y}^{-1} \mathbf{I}$$

But, we likewise know that:

$$\mathbf{V} = \mathcal{Z} \mathbf{I}$$

By comparing the two previous expressions, we can conclude:

$$\mathcal{Z} = \mathcal{Y}^{-1} \quad \text{and} \quad \mathcal{Z}^{-1} = \mathcal{Y}$$