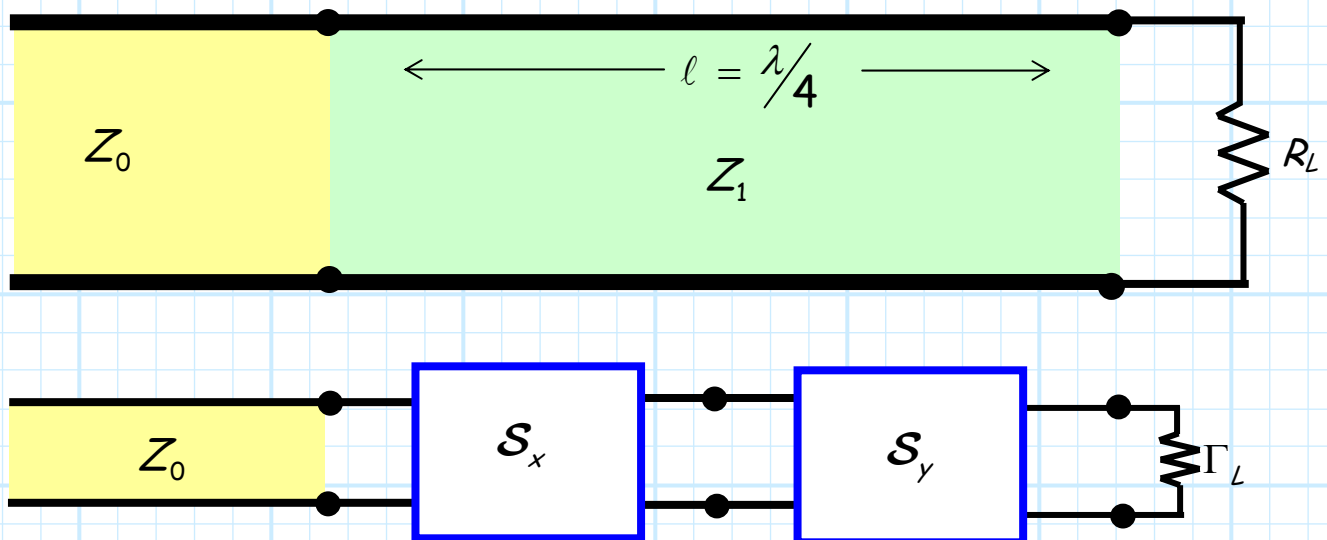


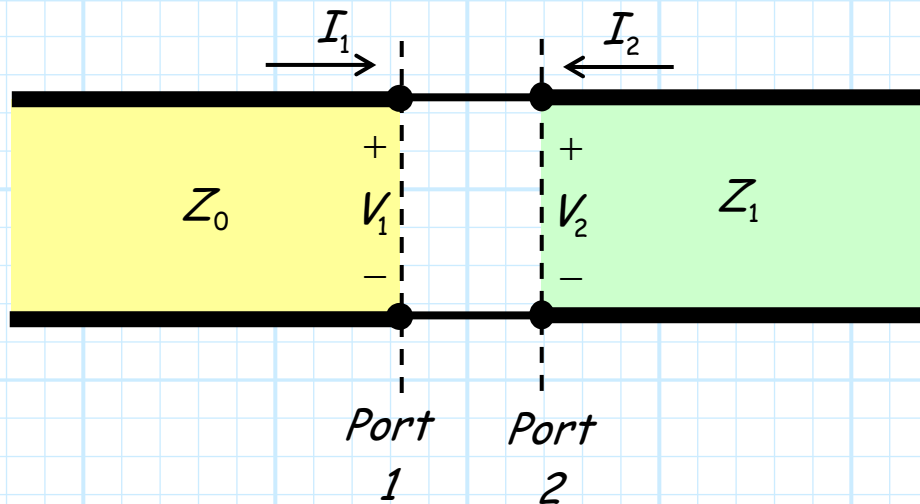
# The Signal Flow Graph of a Quarter-Wave Transformer

A quarter wave transformer can be thought of as a cascaded series of two two-port devices, terminated with a load  $R_L$ :



**Q:** *Two two-port devices? It appears to me that a quarter-wave transformer is **not** that complex. What **are** the two two-port devices?*

**A:** The first is a "connector". Note a connector is the interface between one transmission line (characteristic impedance  $Z_0$ ) to a second transmission line (characteristic impedance  $Z_1$ ).



Recall that we **earlier** determined the scattering matrix of this two-port device:

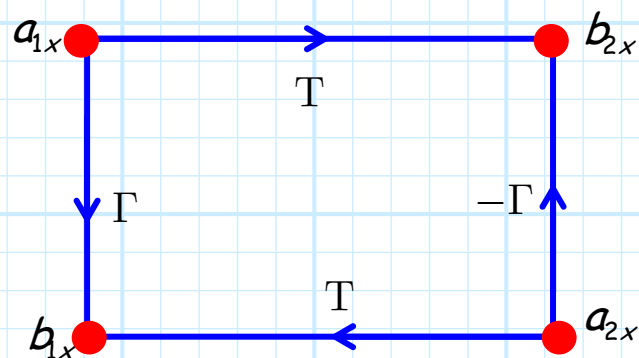
$$S_x = \begin{bmatrix} \frac{Z_1 - Z_0}{Z_1 + Z_0} & \frac{2\sqrt{Z_0 Z_1}}{Z_0 + Z_1} \\ \frac{2\sqrt{Z_0 Z_1}}{Z_0 + Z_1} & \frac{Z_0 - Z_1}{Z_0 + Z_1} \end{bmatrix}$$

This result can be more **compactly** stated as:

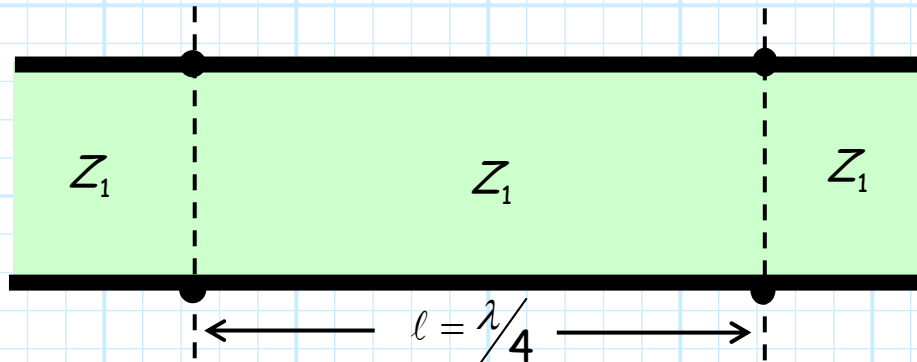
$$S = \begin{bmatrix} \Gamma & T \\ T & -\Gamma \end{bmatrix}$$

where  $\Gamma \doteq \frac{Z_1 - Z_0}{Z_1 + Z_0}$  and  $T \doteq \frac{2\sqrt{Z_0 Z_1}}{Z_0 + Z_1}$

The signal flow graph of this device is therefore:

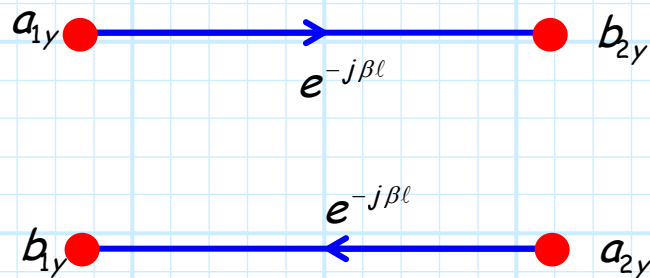


Now, the second two-port device is a quarter wavelength of transmission line.

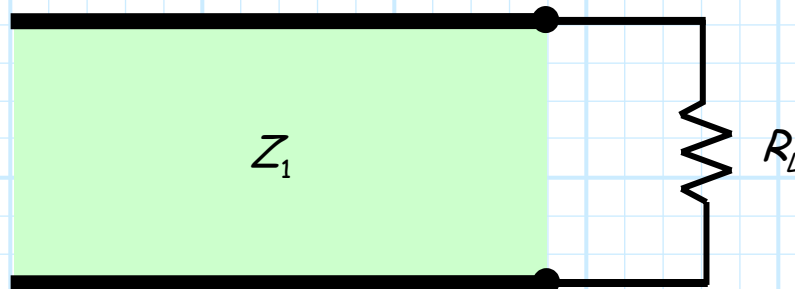


We know that it has the scattering matrix:

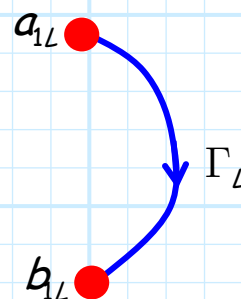
$$S_y = \begin{bmatrix} 0 & e^{-j\beta l} \\ e^{-j\beta l} & 0 \end{bmatrix}$$



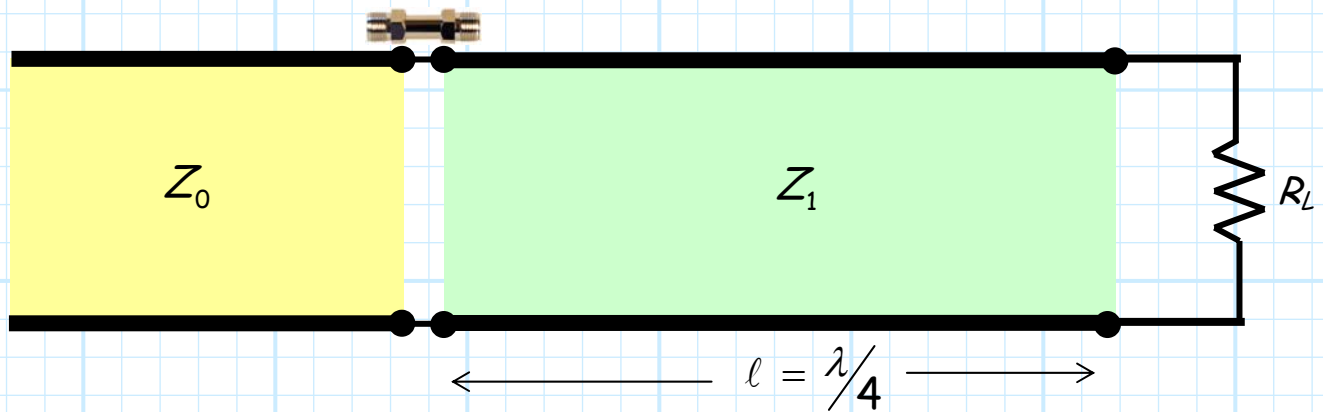
Finally, a load has a "scattering matrix" of:



$$S = \left[ \frac{R_L - Z_1}{R_L + Z_1} \right] = \Gamma_L$$



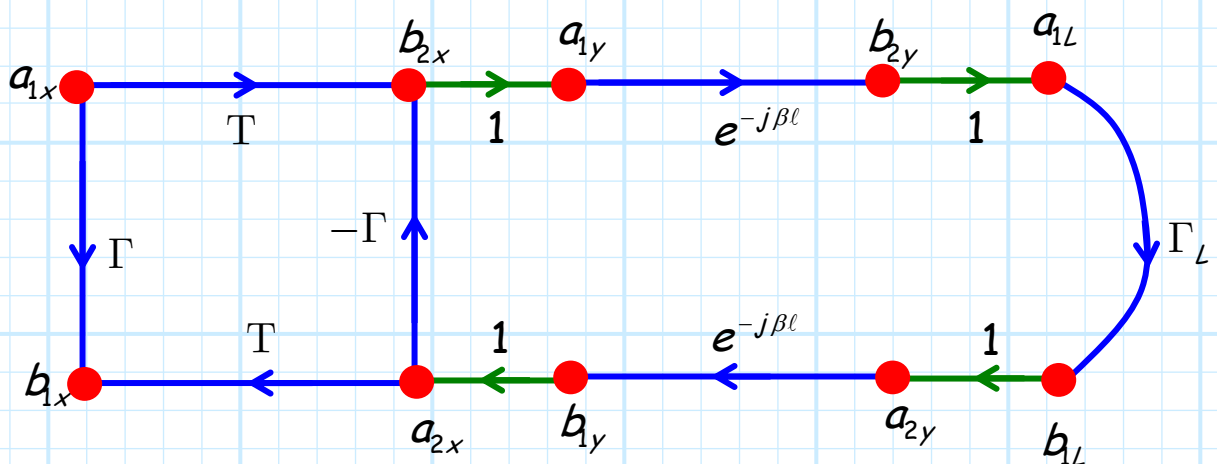
Of course, if we connect the ideal connector to a quarter wavelength of transmission line, and terminate the whole thing with load  $R_L$ , we have formed a **quarter wave matching network!**



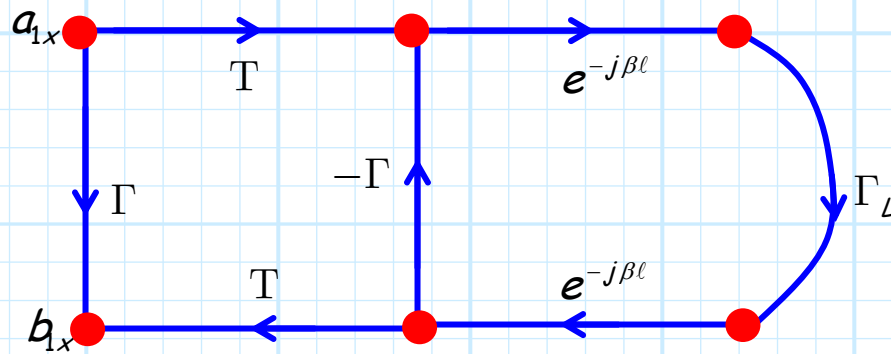
The boundary conditions associated with these connections are likewise:

$$a_{1y} = b_{2x} \quad a_{2x} = b_{1y} \quad a_{1L} = b_{2y} \quad a_{2y} = b_{1L}$$

We can thus put the signal-flow graph pieces together to form the **signal-flow graph** of the quarter wave network:



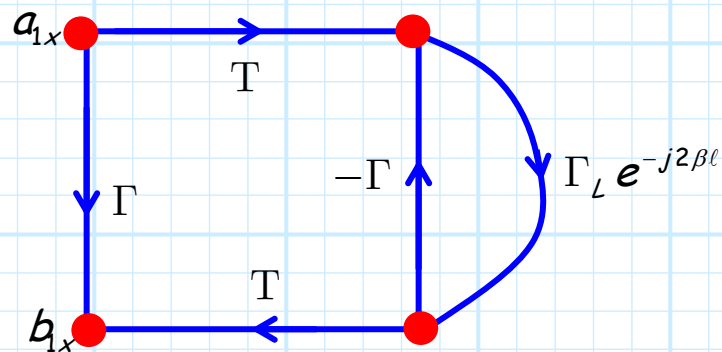
And simplifying:



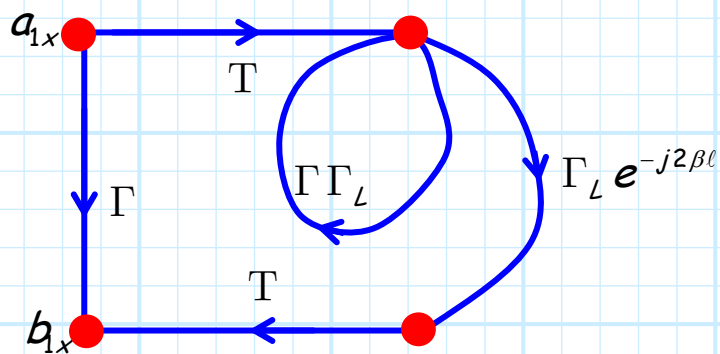
Now, let's see if we can **reduce** this graph to determine:

$$\Gamma_{in} \doteq \frac{b_{1x}}{a_{1x}}$$

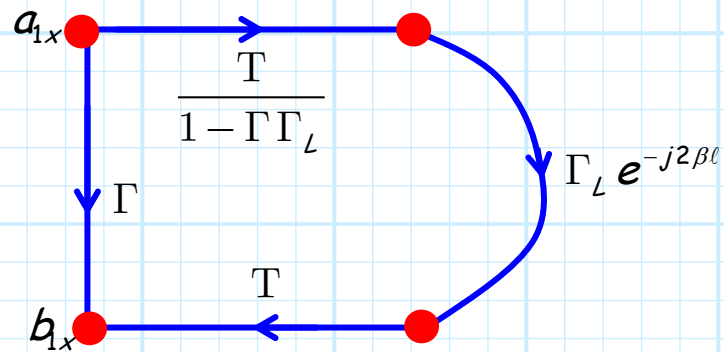
From the **series** rule:



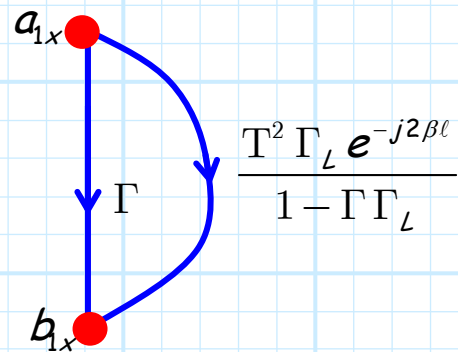
From the **splitting** rule:



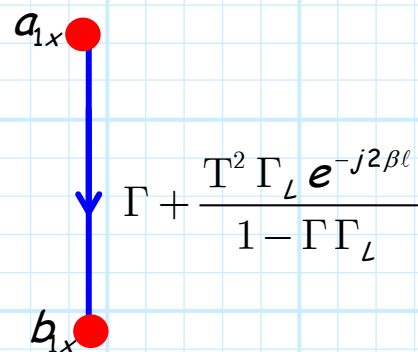
From the **self-loop** rule:



Again with the **series** rule:



And finally with the **parallel** rule:



So that:

$$\Gamma_{in} \doteq \frac{b_{1x}}{a_{1x}} = \Gamma + \frac{T^2 \Gamma_L e^{-j2\beta\ell}}{1 - \Gamma\Gamma_L}$$

**Q:** Hey wait! If the quarter-wave transformer is a **matching network**, shouldn't  $\Gamma_{in} = 0$ ??

**A:** Who says it isn't! Consider now **three important facts**.

For a **quarter wave transformer**, we set  $Z_1$  such that:

$$Z_1^2 = Z_0 R_L \quad \Rightarrow \quad Z_0 = Z_1^2 / R_L$$

**Inserting** this into the scattering parameter  $S_{11}$  of the connector, we find:

$$\Gamma = \frac{Z_1 - Z_0}{Z_1 + Z_0} = \frac{Z_1 - Z_1^2/R_L}{Z_1 + Z_1^2/R_L} = \frac{R_L - Z_1}{R_L + Z_1}$$

Look at this result! For the quarter-wave transformer, the **connector  $S_{11}$  value** (i.e.,  $\Gamma$ ) is the **same** as the **load reflection coefficient  $\Gamma_L$** :

$$\Gamma = \frac{R_L - Z_1}{R_L + Z_1} = \Gamma_L \quad \leftarrow \text{Fact 1}$$

Since the connector is **lossless** (unitary scattering matrix!), we can conclude (and likewise show) that:

$$1 = |S_{11}|^2 + |S_{21}|^2 = |\Gamma|^2 + |T|^2$$

Since  $Z_0$ ,  $Z_1$ , and  $R_L$  are all real, the values  $\Gamma$  and  $T$  are also **real valued**. As a result,  $|\Gamma|^2 = \Gamma^2$  and  $|T|^2 = T^2$ , and we can likewise conclude:

$$\Gamma^2 + T^2 = 1 \quad \leftarrow \text{Fact 2}$$

Likewise, the Z1 transmission line has  $\ell = \lambda/4$ , so that:

$$2\beta\ell = 2\left(\frac{2\pi}{\lambda}\right)\frac{\lambda}{4} = \pi$$

where you of course recall that  $\beta = 2\pi/\lambda$ ! Thus:

$$e^{-j2\beta\ell} = e^{-j\pi} = -1 \quad \leftarrow \text{Fact 3}$$

As a result:

$$\Gamma_{in} = \Gamma + \frac{T^2 \Gamma_L e^{-j2\beta\ell}}{1 - \Gamma\Gamma_L} = \Gamma - \frac{T^2 \Gamma_L}{1 - \Gamma\Gamma_L}$$

And using the **newly discovered** fact that (for a correctly designed transformer)  $\Gamma_L = \Gamma$ :

$$\Gamma_{in} = \Gamma - \frac{T^2 \Gamma_L}{1 - \Gamma\Gamma_L} = \Gamma - \frac{T^2 \Gamma}{1 - \Gamma^2}$$

And also are **recent discovery** that  $T^2 = 1 - \Gamma^2$ :

$$\Gamma_{in} = \Gamma - \frac{T^2 \Gamma}{1 - \Gamma^2} = \Gamma - \frac{T^2 \Gamma}{T^2} = 0$$

**A perfect match!** The quarter-wave transformer does indeed work!