

# The Scattering Matrix

At “**low**” frequencies, we can completely characterize a **linear** device or network using an **impedance** matrix, which relates the currents and voltages at **each** device terminal to the currents and voltages at **all** other terminals.

But, at microwave frequencies, it is **difficult** to measure currents and voltages!

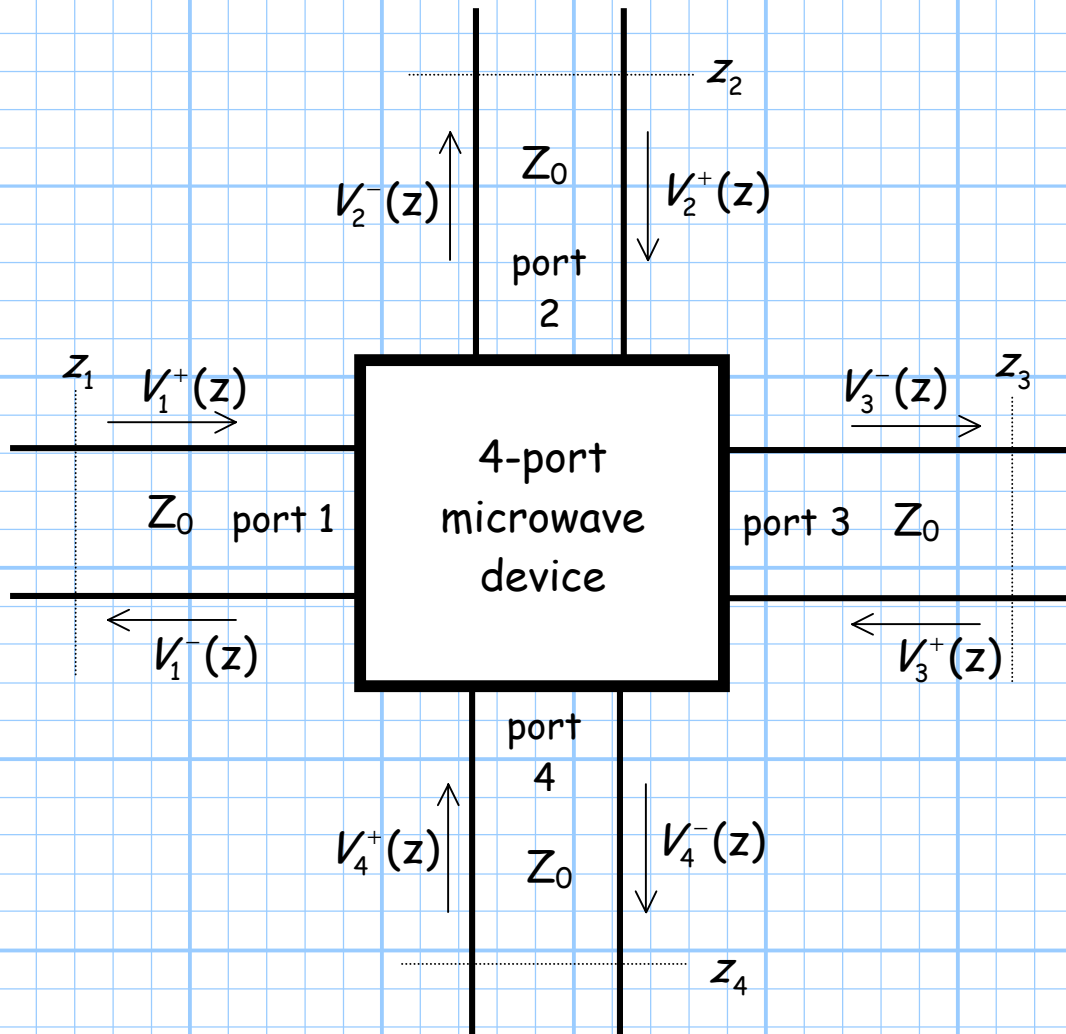


\* Instead, we can measure the **incident** and **reflected** waves  $V_0^+(z)$  and  $V_0^-(z)$ .

\* In other words, we can determine the relationship between the incident and reflected wave at **each** device terminal to the incident and reflected waves at **all** other terminals.

These relationships are completely represented by the **scattering matrix**. It **completely** describes the behavior of a linear, multi-port device at frequency  $\omega$ .

Consider the **4-port** microwave device shown below:



Say we know that there exists an **incident** wave on **port 1** (i.e.,  $V_1^+ \neq 0$ ), while the incident waves on all other ports are known to be **zero** (i.e.,  $V_2^+ = V_3^+ = V_4^+ = 0$ ). Say we then measure the wave **out** of **port 2** (i.e., determine  $V_2^-$ ). The complex ratio between  $V_1^+(z = z_1)$  and  $V_2^-(z = z_2)$  is known as the **scattering parameter**  $S_{21}$ :

$$S_{21} = \frac{V_2^-(z = z_2)}{V_1^+(z = z_1)} = \frac{V_2^- e^{+j\beta z_2}}{V_1^+ e^{-j\beta z_1}} = \frac{V_2^-}{V_1^+} e^{+j\beta(z_2 + z_1)}$$

Likewise, the scattering parameters  $S_{31}$  and  $S_{41}$  are:

$$S_{31} = \frac{V_3^-(z = z_3)}{V_1^+(z = z_1)} \quad \text{and} \quad S_{41} = \frac{V_4^-(z = z_4)}{V_1^+(z = z_1)}$$

We of course could also define, say, scattering parameter  $S_{34}$  as the ratio between the complex values  $V_4^+(z = z_4)$  (describing the wave **into** port 4) and  $V_3^-(z = z_3)$  (describing the wave **out of** port 3), given that the input to all other ports (1,2, and 3) are zero.

Thus, more **generally**, the ratio of the wave incident on port  $n$  to the wave emerging from port  $m$  is:

$$S_{mn} = \frac{V_m^-(z = z_m)}{V_n^+(z = z_n)} \quad (\text{given that } V_k^+ = 0 \text{ for all } k \neq n)$$

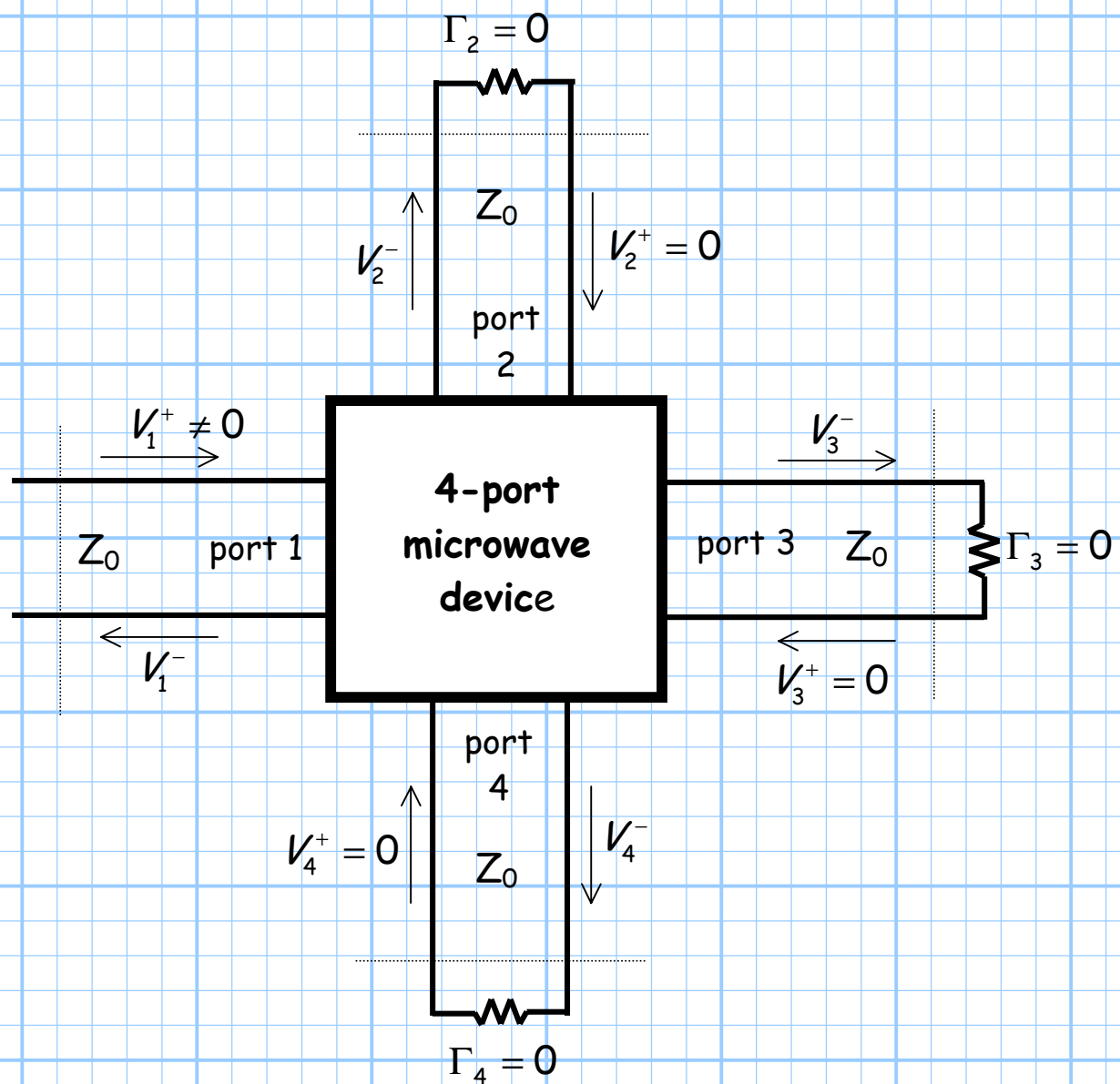
Note that frequently the port boundary locations are assigned a zero value (e.g.,  $z_1 = 0$ ,  $z_2 = 0$ ). This of course simplifies the scattering parameter calculation:

$$S_{mn} = \frac{V_m^-(z_m = 0)}{V_n^+(z_n = 0)} = \frac{V_m^- e^{+j\beta 0}}{V_n^+ e^{-j\beta 0}} = \frac{V_m^-}{V_n^+}$$

We will generally assume that the port locations are defined as  $z = 0$ , and thus use the above notation. But remember where this expression came from!

**Q:** How can we make sure that **only one** incident wave is non-zero ?

**A:** **Terminate** all ports with a **matched load**!



If the ports are terminated in a **matched load** (i.e.,  $Z_L = Z_0$ ), then  $\Gamma_n = 0$  and therefore:

$$V_n^+ = \Gamma_n V_n^- = 0$$

In other words, terminating a port ensures that there will be **no signal** incident on that port!

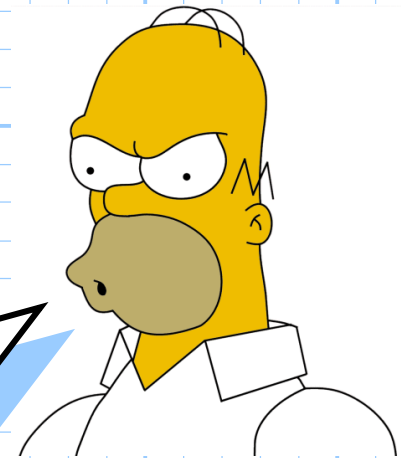
Wait a minute—you've messed this all up! **Earlier** you said that:

$$V_n^- = \Gamma_n V_n^+$$

but **now** you say that:

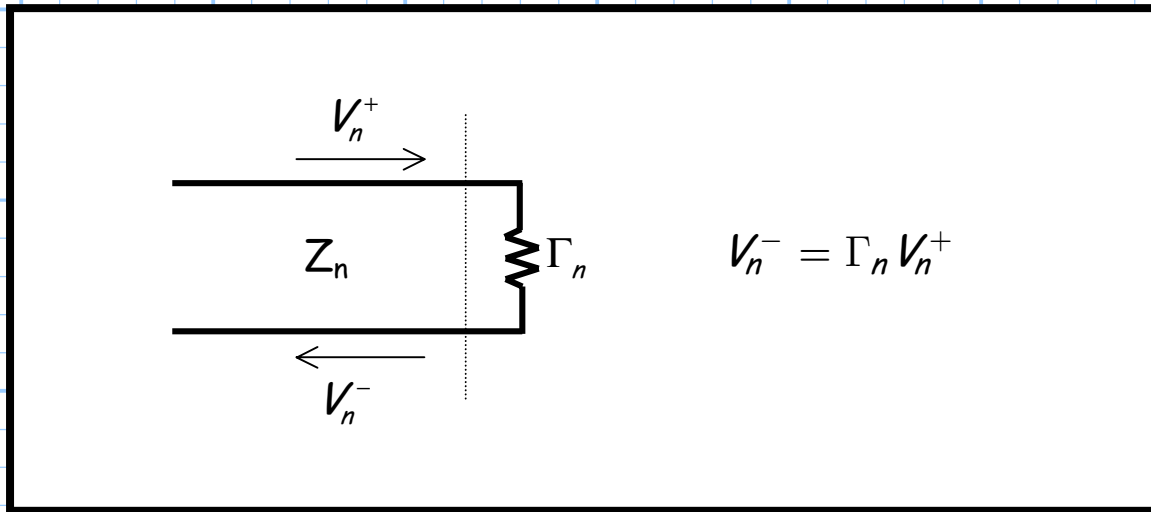
$$V_n^+ = \Gamma_n V_n^-$$

Make up your mind! **Which one** is correct?



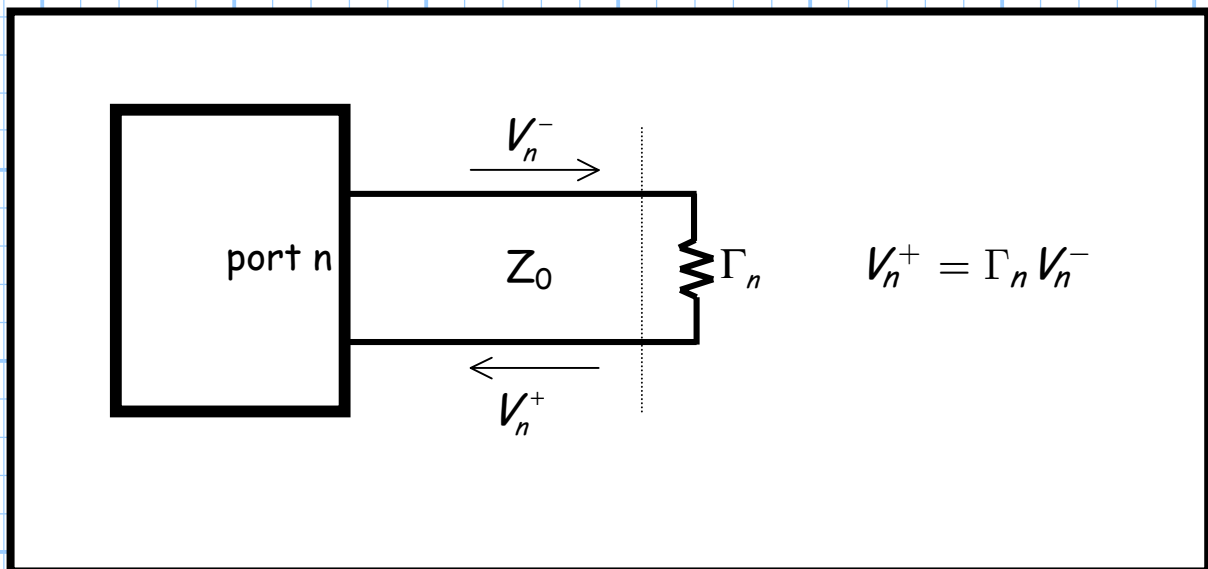
Actually, **both** statements are correct! You must be careful to understand the **physical definitions** of the constants  $V_n^+$  and  $V_n^-$ .

For example, we **originally** analyzed this case:



In this case, the complex constant describing the wave **incident** on the load is  $V_n^+$  and the constant describing the **reflected** wave is  $V_n^-$ .

Constrast this with the case we are **now** considering:



Here, we defined the constant of the wave incident on **port  $n$**  as  $V_n^+$ . As a result, the constant  $V_n^+$  now describes the wave **reflected** from the load. Likewise, the wave **incident** on the load is now described by constant  $V_n^-$ .

Perhaps we could more generally state that:

$$V^{\text{reflected}} = \Gamma V^{\text{incident}}$$



For each case, **you** must be able to correctly identify the constants describing the wave **incident** on, and **reflected** from, some load.

Like most equations in engineering, the **variable names** can **change**, but the **physics** described by the mathematics will **not**!

Now, **back** to our discussion of **S-parameters**. We found that:

$$V_m^- = S_{mn} V_n^+ \quad (\text{given that } V_k^+ = 0 \text{ for all } k \neq n)$$

Which we can now **equivalently** state as:

$$V_m^- = S_{mn} V_n^+ \quad (\text{given that all ports } k \neq n \text{ are matched})$$

Say that we have waves **simultaneously** incident on each of the four ports of our device.

**Q:** *What then is the output at, for example, port 3??*

**A:** Use **superposition** !!

Since the device is **linear**, the output at port 3 due to **all** the incident waves is simply the coherent **sum** of the output at port 3 due to **each** wave:

$$V_3^- = S_{34} V_4^+ + S_{33} V_3^+ + S_{32} V_2^+ + S_{31} V_1^+$$

Or, more generally, the output at port  $m$  of an  $N$ -port device is:

$$V_m^- = \sum_{n=1}^N S_{mn} V_n^+$$

This expression can be written in matrix form as:

$$\bar{V}^- = \bar{S} \bar{V}^+$$



Where  $\bar{\mathbf{V}}^-$  is the vector:

$$\bar{\mathbf{V}}^- = [V_1^-, V_2^-, V_3^-, \dots, V_N^-]^T$$

and  $\bar{\mathbf{V}}^+$  is the vector:

$$\bar{\mathbf{V}}^+ = [V_1^+, V_2^+, V_3^+, \dots, V_N^+]^T$$

Therefore  $\bar{\bar{\mathbf{S}}}$  is the **scattering matrix**:

$$\bar{\bar{\mathbf{S}}} = \begin{pmatrix} S_{11} & \cdots & S_{1n} \\ \vdots & \ddots & \vdots \\ S_{m1} & \cdots & S_{mn} \end{pmatrix}$$

The scattering matrix is a  $N$  by  $N$  matrix that **completely characterizes** a linear,  $N$ -port device (at frequency  $\omega$ ).