11.2 Stability

Reading Assignment: pp. 542-548

A gain element is an active device. One potential problem with every active circuit is its stability

HO: STABILITY

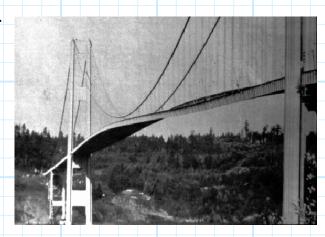
Jim Stiles The Univ. of Kansas Dept. of EECS

Stability

Q: So all there is to making a good microwave amplifier is the design of proper matching networks?

A: There is one other problem that confronts the microwave amplifier designer. That problem is **stability** (of the amplifier, not the designer).

An unstable amplifier is also known as an oscillator—a source of microwave energy!



Q: Under what conditions will an amplifier oscillate?

A: An amplifier will go unstable if either of these two conditions are true:

$$\left|\Gamma_{out}\right| = \left|\mathcal{S}_{22} + \frac{\mathcal{S}_{12}\mathcal{S}_{21}\Gamma_{s}}{1 - \mathcal{S}_{11}\Gamma_{s}}\right| > 1.0$$

$$\left|\Gamma_{in}\right| = \left|\mathcal{S}_{11} + \frac{\mathcal{S}_{12}\mathcal{S}_{21}\Gamma_{L}}{1 - \mathcal{S}_{22}\Gamma_{L}}\right| > 1.0$$

In other words, the amplifier will oscillate if either the input or output reflection coefficient of the gain element has a magnitude greater than one.

Q: Hey wait! I thought we learned that the **maximum** value of any reflection coefficient magnitude was 1 (i.e., $|\Gamma| \le 1.0$)—this defined the **validity region** of our Smith Chart!

A: Remember, the inequality $|\Gamma| \le 1.0$ is true for any passive load or device. Our gain element is an active device—it must have a DC source of power.

As a result, we find that $|\Gamma| > 1.0$ is quite **possible!**

Q: But, we learned that the region outside the $|\Gamma| = 1.0$ circle on the Smith Chart corresponded to loads with negative values of resistance. Does this mean that Z_{in} or Z_{out} could have real (i.e. resistive) components that are negative?

A: That's exactly what it means!

Q: What is a negative resistor exactly?

A: Ohm's law still applies—negative resistance or not. As a result, we would find for a $-10~\Omega$ resistor that:

$$\frac{V}{T} = -10$$
 \Rightarrow $V = -10 I$

G: 555

A: The result above simply means that the current through a negative resistor is 180° out-of-phase with the voltage across it.

→ The resistor current is at its minimum value when the voltage across it is at it maximum—and vice versa!

This behavior drives our amplifier circuit a little wacky, and it begins to oscillate!

Q: So how do we avoid this unfortunate occurrence?

A: Recall that amplifier instability occurs when:

$$\left|\Gamma_{out}\right| = \left|\mathcal{S}_{22} + \frac{\mathcal{S}_{12}\mathcal{S}_{21}\Gamma_{s}}{1 - \mathcal{S}_{11}\Gamma_{s}}\right| > 1.0$$

$$\left|\Gamma_{in}\right| = \left|\mathcal{S}_{11} + \frac{\mathcal{S}_{12}\mathcal{S}_{21}\Gamma_{L}}{1 - \mathcal{S}_{22}\Gamma_{L}}\right| > 1.0$$

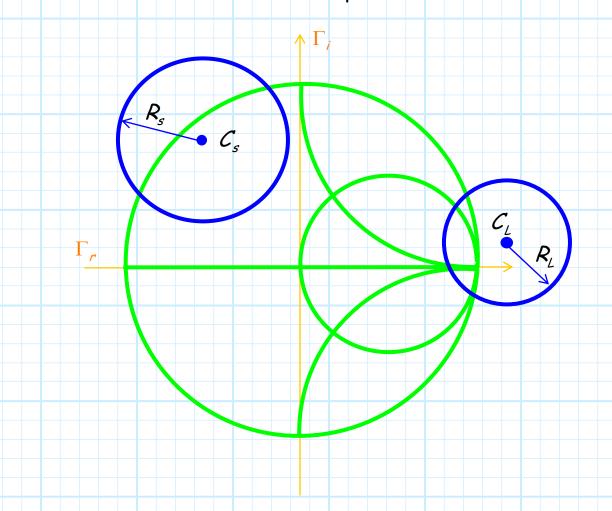
Thus, for a given gain element (i.e., S_{11} , S_{21} , S_{22} , S_{12}), amplifier stability is determined by the value of Γ_{L} and Γ_{s} .

We can **solve** the above equations to determine the specific range of values of Γ_{L} and Γ_{s} that will **induce oscillation**. The results are provided in page 543 and 544 of your text.

We find that these unstable values—when plotted on the complex Γ plane—form a circle. These circles are know as a stability circles.

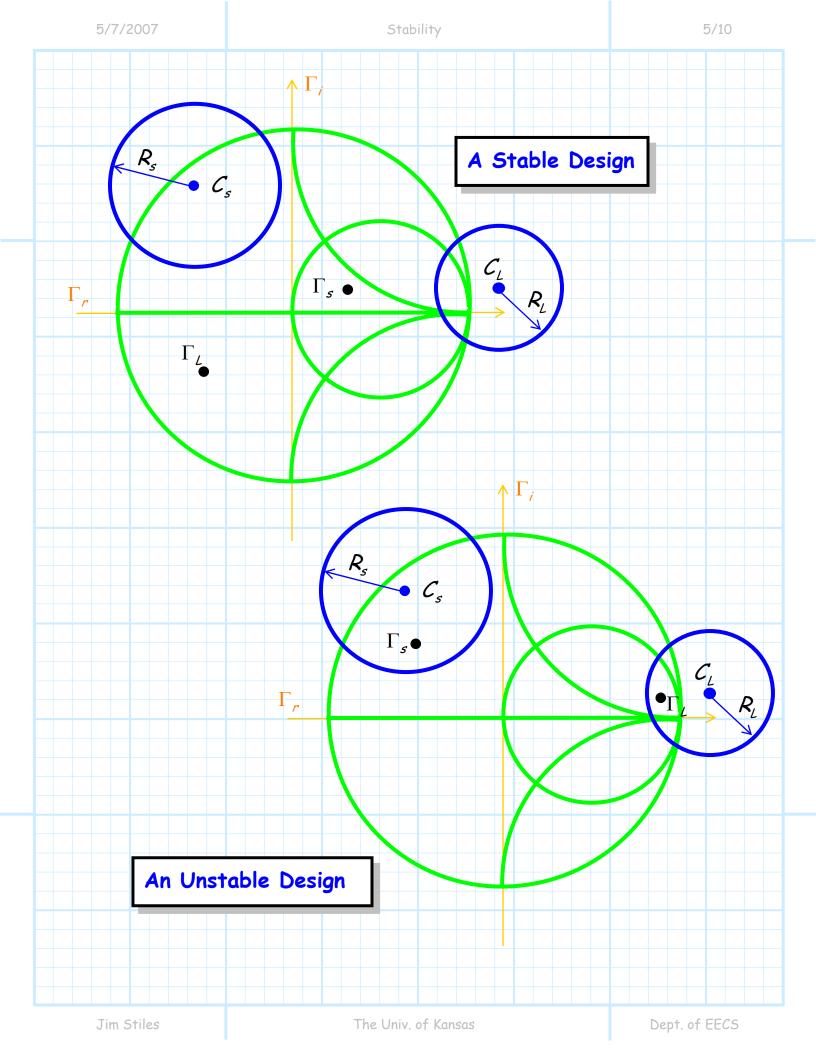
These gain circles are defined as a complex value C, which specifies the location of the stability circle **center** on the complex Γ plane, and a real value R, which specifies the **radius** of the stability circle.

There is **one** stability circle for Γ_{L} (i.e., C_{L} and R_{L}) and **another** for Γ_{s} (i.e., C_{s} and R_{s}). Typically, the Γ values that lie **inside** the circle will create amplifier oscillation.



Q: So what do we use these stability circles for?

A: As an amplifier designer, we must make sure that our design values Γ_L and Γ_s lie outside these circles—otherwise, our well-designed amplifiers will oscillate!

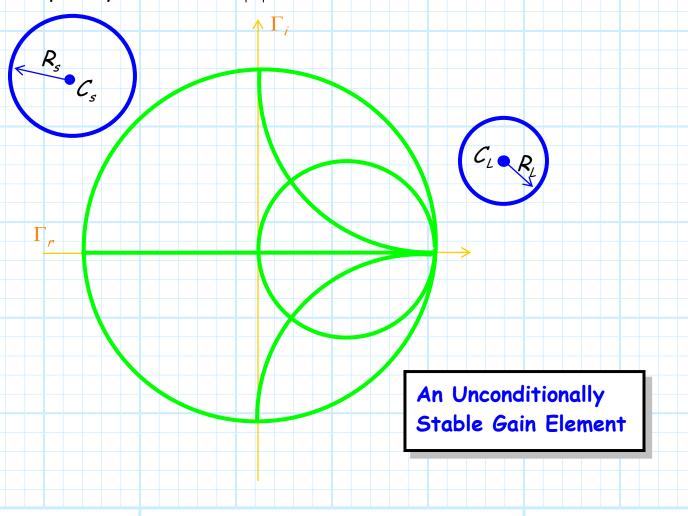


Q: Yikes! Must we always determine these circles and check our design for instability?

A: Not necessarily! Some gain elements are unconditionally stable. As the name suggests, these gain elements result in stable amplifiers for any and all realizable values of Γ_{ℓ} and Γ_{s} .

Q: So an unconditionally stable gain element has stability circles with zero radius (i.e., R = 0)?

A: Could be, but all that is required for a gain element to be unconditionally stable is for its stability circles to lie completely outside the $|\Gamma|=1$ circle.



For this condition, we find that the values of Γ_{ℓ} and Γ_{s} that result in an unstable amplifier must have a magnitude **greater** than 1 (i.e., $|\Gamma_{\ell}| > 1$ or $|\Gamma_{s}| > 1$).

But, we can presume that our the loads and sources attached to our amplifier will always have positive resistances, such that $|\Gamma_{L}| < 1$ and $|\Gamma_{s}| < 1$.

Thus, an amplifier constructed with an unconditionally stable gain element will likewise be unconditionally stable!

Q: How will I recognize an unconditionally stable gain element if I see one? Must I determine and plot the stability circles?

A: There are **tests** that we can apply—using the scattering parameters S_{11} , S_{21} , S_{22} , S_{12} —to more **directly** determine if a gain element is unconditionally stable.

First, we find some necessary conditions for a gain element to be unconditionally stable are:

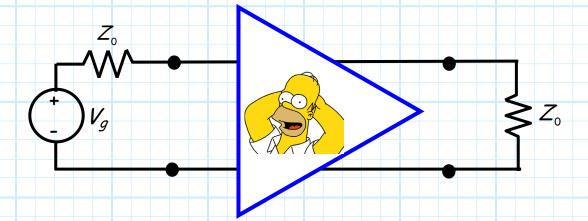
$$|S_{11}|^2 < 1$$
 and $|S_{22}|^2 < 1$

If our gain element is **unilateral** (i.e., $S_{12} = 0$ or approx. $|S_{12}| \ll |S_{21}|$) then the above conditions are likewise **sufficient** for unconditional stability.

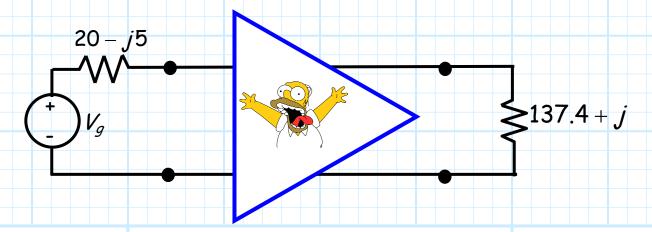
Otherwise, (for $S_{12} \neq 0$) we find that our gain element must pass **two more tests**—tests that are provided on page 545 of your text.

Q: Do we really **care** if our design is unconditionally stable? Aren't we really **just** concerned with whether our design values Γ_L and Γ_s lie inside the stability circle?

A: Remember, the values Γ_{L} and Γ_{s} are determined for the specific values of source and load impedances connected to the amplifier (presumably Z_{0}).



But what if the resulting amplifier is **not** connected to these ideal sources? The ideal source or load impedance Z_0 is **never** achieved with **perfection**, and often achieved **not at all** (consider all the **narrow-band** devices we have studied!).



Thus, since we do not specifically know what source and load impedances our amplifier might encounter, we had generally design an amplifier that is stable for them all—one that's unconditionally stable!

Q: Anything else we need to know about amplifier stability?

A: One last very important thing.

Recall that amplifiers, like all microwave devices, are **dependent on frequency**. Thus, all of the important values involved in our design (e.g., Γ_s , S_{11} , S_{21} , S_{22} , S_{12} , Γ_o) will **change** as a function of frequency!

Q: I see, amplifier performance, most notably gain, will change as a function of frequency, and so maximum power transfer will occur at just our design frequency.

We've seen this kind of thing before!

A: True, but for amplifiers there is also a new twist.

The amplifier stability conditions (i.e., stability tests) must be satisfied at any and all frequencies!

If for even one frequency we find that either:

$$\left|\Gamma_{out}\right| = \left|S_{22} + \frac{S_{12}S_{21}\Gamma_{s}}{1 - S_{11}\Gamma_{s}}\right| > 1.0 \quad \text{or} \quad \left|\Gamma_{in}\right| = \left|S_{11} + \frac{S_{12}S_{21}\Gamma_{L}}{1 - S_{22}\Gamma_{L}}\right| > 1.0$$

then our amplifier will oscillate—even if that frequency is not our "design frequency"!

This makes amplifier stability a much more significant and difficult problem than you might otherwise think.

→ An unconditionally stable amplifier must be unconditionally stable at all frequencies!

Jim Stiles The Univ. of Kansas Dept. of EECS